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# Detecting Small Size Mass Loading Using Transversely Coupled SAW Resonator

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**Abstract:** In the detection of small size mass loading, such as a single cell, a micro droplet or an aerosol particle, the sensors with longitudinally coupled surface acoustic wave resonator (LC-SAWR) structure can hardly avoid waveform distortions. The relative size of mass loading to the sensitive surface of the detector is the main factor affecting the response of LC-SAWR. The smaller the relative size, the worse the waveform distortion. In order to avoid influences from the mass loading's size, in this paper, a transversely coupled SAW resonator (TC-SAWR) was proposed in order to achieve high performance in sensing small size mass loadings. For the design and simulation of TC-SAWR, the two-dimensional coupling of model (2D-COM) theory and finite element method (FEM) were used in this work. In the experiment,  $SiO_2$  was deposited on the sensor's surface as a small size mass loading. The results from simulation and experiment mutually demonstrated the advantage of TC-SAWR to conquer waveform distortion in the detection of small size mass loading.

**Keywords:** surface acoustic wave; small size load; biochemical sensors; waveform distortion; 2D-COM theory; FEM Simulation

## 1. Introduction

The surface acoustic wave resonator (SAWR) was used mainly as an extremely narrowband filter until Martin reported its use to measure chemical vapors [1]. Since then, SAWRs have been widely used in the field of sensing for their advantages such as low loss, high sensitivity and stability. In recent years, SAWRs have received more attention in the field of biosensing, especially in the measurement of small mass loading [2,3]. According to the resonance mode, SAWR could be sorted as a longitudinally coupled SAW resonator (LC-SAWR) and transversely coupled resonator (TC-SAWR). Normally, researchers take LC-SAWRs as sensor chips such as vapor detector [4], mass microbalance [5], PM2.5 detector [6], strain sensor [7,8] and temperature sensors [9]. In these cases, the physical quantities measured were uniformly exerted on the surface of the acoustic wave device. However, in biological detection, the substances measured, such as cells and proteins, are of small size; thus, it is difficult for these substances to completely and uniformly cover the surface of the detector [10–13]. In our previous work, we have discussed the effect of non-uniformly distributed loads on the LC-SAWR's response, which showed waveform distortions caused by small size mass loadings [14]. In order to achieve high performance in the sensing of small size mass loading, a TC-SAWR was proposed in this paper as a detector in order to conquer waveform distortion.

When a LC-SAWR is working at its resonate mode, the device could be modeled as a combination of several independent channels [14]. If the load does not fully cover the device's aperture, it can only be applied to part of the channels. Then, the overall response of the device would be composed of the loaded channels' responses and the



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non-loaded channels' responses. The center frequency of the loaded channel is lower than that of the no-loaded channel, resulting in waveform distortion. (Figure 1). For TC-SAWR, the acoustic wave resonates transversely in the aperture, forming resonance peaks of each mode (Figure 2). In this case, the device cannot be divided into multiple independent channels and needs to be analyzed integrally. As a result, the load acting on the TC-SAWR will change the resonance frequency of each mode, regardless of whether it covers the aperture completely or not, which avoids waveform distortion caused by the small size load.

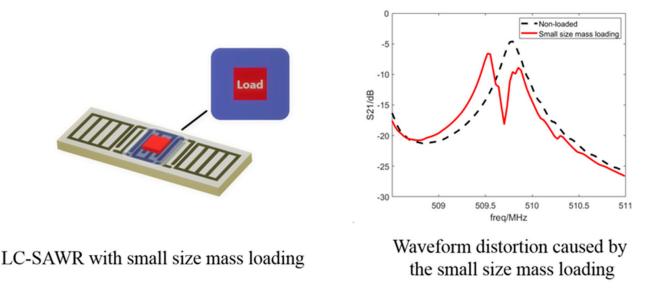


Figure 1. LC-SAWR with small size mass loading (left) and the waveform distortion (right).

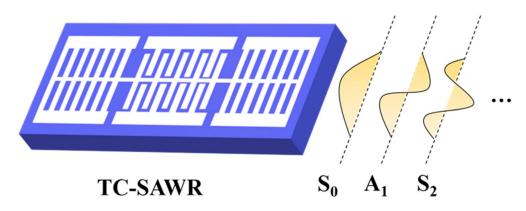


Figure 2. Basic structure of TC-SAWR and its transverse resonance modes.

In the past, the one-dimensional coupled mode theory (1D-COM) has been used for the simulation and design of TC-SAWRs [15,16]. However, this method can only compute a limited number of transverse modes; thus, it is not accurate enough to simulate the response of TC-SAWR. In addition, the 1D-COM theory is difficult with respect to modeling and simulating the device with small size loads. To solve these problems, we used the two-dimensional coupling-of-modes theory (2D-COM) and the finite element method (FEM) in order to obtain a more accurate simulation of TC-SAWR and to calculate the influence of small size loads on it.

The experiment was implemented in this work to verify the simulation result. In the experiment, TC-SAWR with a  $SiO_2$  cylinder attached to its surface was fabricated and measured.

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In this work, the effect of small size load attached to the TC-SAWR sensors were studied theoretically and experimentally. They consistently showed that the frequency response curve of TC-SAWR had no distortion under small size load, which illustrated the advantage of TC-SAWR in the field of biochemical detection.

#### 2. Materials and Methods

## 2.1. Two-Dimensional Coupling-of-Modes Theory

The two-dimensional coupling-of-modes theory (2D-COM) is a simulation method of surface acoustic wave devices by considering the diffraction effect. It was originally proposed by Haus to analyze the waveguide effect (transverse mode) of surface acoustic wave devices [17]. The TC-SAWR works in the transverse resonant modes; thus, the 2D-COM theory is very suitable for its simulation. Recently, Xiao et al. proposed a new general biquadratic form of 2D-COM equations, which further improved the simulation accuracy of this method [18]. In this paper, the 2D-COM equations of biquadratic form were used. The equations are shown in Equation (1).

$$\begin{cases} \frac{\partial u_{1}}{\partial x} = -jk_{0}u_{1} + j\frac{\gamma_{1}}{2k_{0}}\frac{\partial^{2}u_{1}}{\partial y^{2}} + j\frac{\gamma_{2}}{24k_{0}^{3}}\frac{\partial^{4}u_{1}}{\partial y^{4}} + j\kappa u_{2} + j\alpha V \\ \frac{\partial u_{2}}{\partial x} = jk_{0}u_{2} - j\frac{\gamma_{1}}{2k_{0}}\frac{\partial^{2}u_{2}}{\partial y^{2}} - j\frac{\gamma_{2}}{24k_{0}^{3}}\frac{\partial^{4}u_{2}}{\partial y^{4}} - j\kappa^{*}u_{1} - j\alpha^{*}V \\ \frac{\partial^{2}I}{\partial x\partial y} = -2j\alpha^{*}u_{1} - 2j\alpha u_{2} + j\omega CV \end{cases}$$
(1)

Here,  $u_1$  and  $u_2$  stands for the particle displacements of waves propagating forward and backward directions, respectively; and  $k_0$  is the wave number according to the electric period of the IDT.  $\gamma_1$  and  $\gamma_2$  are the parameters for characterizing the anisotropy of the piezoelectric substrate;  $\kappa$ ,  $\alpha$ , C, V and I represent coupling coefficient, excitation coefficient, static capacitance, input voltage and current.

The parameters  $\gamma_1$  and  $\gamma_2$  can be obtained by calculating the second and fourth derivatives at the zero point of the slowness curve [18,19]. The slowness curves of each region of the device with Al/ST-X Quartz structure are shown in Figure 3.

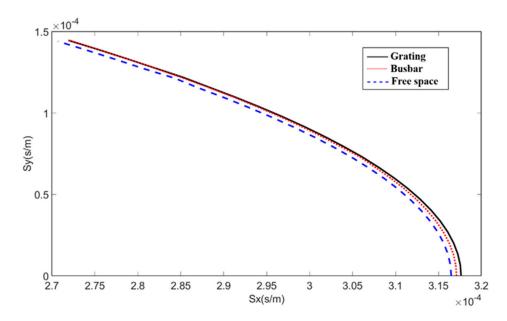


Figure 3. Slowness curves of each region of the TC-SAWR device.

The values of  $\gamma_1$  and  $\gamma_2$  in different regions were further obtained, as shown in Table 1.

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	<b>Table 1.</b> Anisotro	pic parameters $\gamma_1$	$_{1}$ and $\gamma_{2}$ of each region.
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Region	<b>γ</b> 1	γ2
Grating	-1.3799	-0.7284
Busbar	-1.3604	-0.6634
Free space	-1.3928	-0.5000

The parameters such as velocity, coupling coefficient, excitation coefficient and static capacitance were determined by using the finite element method [20], as shown in Table 2.

Table 2. COM parameters.

Parameter	Value
SAW velocity	3148 (m/s)
Normalized reflection coefficient	0.019
Normalized excitation coefficient	$2.48 \times 10^{-5}$
Static capacitance	$3.99 \times 10^{-11} \text{ (F)}$

All the above parameters were close to the values reported in the other literature [8,18] (the subtle differences were caused by factors such as substrate material, device structure and calculation accuracy).

## 2.2. PDE Modeling of Transversely Coupled SAW Resonator

The 2D-COM equations can be numerically solved by the PDE module of COMSOL Multiphysics [21]. Firstly, the geometric model of TC-SAWR was established, as shown in Figure 4. The substrate material and structural parameters of the model are listed in Table 3.

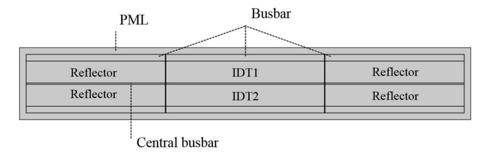


Figure 4. Geometric model of TC-SAWR constructed in COMSOL.

**Table 3.** The substrate material and structural parameters of TC-SAWR.

Substrate Material: ST-X Quartz				
Wavelength (λ)	15.6 μm			
Number of IDT <sub>1,2</sub> fingers	120 pairs			
Number of reflecting fingers	105 pairs			
Finger width	3.9 μm			
Finger height	200 nm			
Width of central busbar	15.6 μm			
Aperture	249.6 μm			

Here, IDT1 and IDT2 are the input and output ports, respectively.

After the geometric model is meshed properly, the 2D-COM equations would be solved by the finite element method. Figure 5 demonstrates the calculated transverse modes of order 1~4, respectively.

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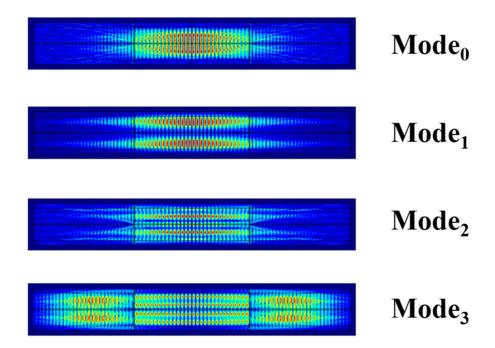


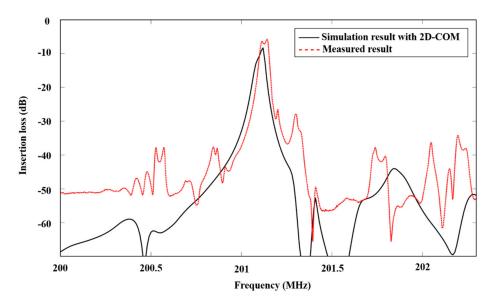
Figure 5. Transverse modes of order 1~4.

The port current should be calculated by the following integral.

$$I = \iint_{Region} -2j\alpha^* u_1 - 2j\alpha u_{2-} + j\omega CV$$
 (2)

The description of variables in Equation (2) is written above (line 86 to 90).

Once the port current is determined, the insertion loss  $(S_{21})$  of TC-SAWR is obtained. The parameter  $S_{21}$  of the fabricated TC-SAWR device was measured by a VNA (Agilent E5071B Network analyzer). The frequency response of  $S_{21}$  is shown in Figure 6. The measured and calculated center frequencies were 201.13 MHz and 201.11 MHz, respectively. The measured result was in good agreement with the simulation.

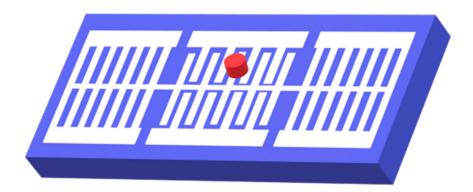


**Figure 6.** Insertion loss  $(S_{21})$  of TC-SAWR calculated by using 2D-COM (black solid curve) and measured (red dotted curve).

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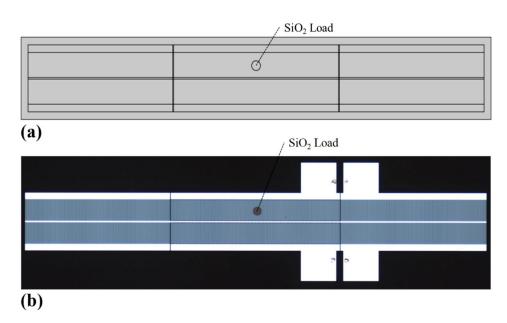
### 3. Result and Discussion

In order to study the effect of small-size mass loading on the response of TC-SAWR device, we designed a  $SiO_2$  cylinder with a diameter of 100  $\mu$ m and a height of 300 nm attached to the device surface, as shown in Figure 7.



**Figure 7.** TC-SAWR device with a SiO<sub>2</sub> cylinder attached to its surface.

The attachment site was located at the central point of IDT1. The simulation model and the photo of fabricated TC-SAWR device are shown in Figure 8.



**Figure 8.** (a) Geometric model of TC-SAWR with SiO<sub>2</sub> load constructed in COMSOL; (b) fabricated TC-SAWR device with SiO<sub>2</sub> load.

Figures 9 and 10 showed the simulated and measured results of frequency response with and without  $SiO_2$  load, respectively. They consistently showed that the frequency response curve of TC-SAWR had no distortion under small size load.

More specifically, under the effect of mass loading, the insertion loss increased around 2 dB, and the center frequency shifted around 14 kHz in the simulation result. On the other hand, the insertion loss increased by 2 dB, and the center frequency shifted around 10 kHz in the measured result. The differences between simulation and experimental results were caused by multiple factors, such as the thickness error of  $SiO_2$  load, the fabrication error of SAW chip and the solution accuracy of the 2-D COM theory.

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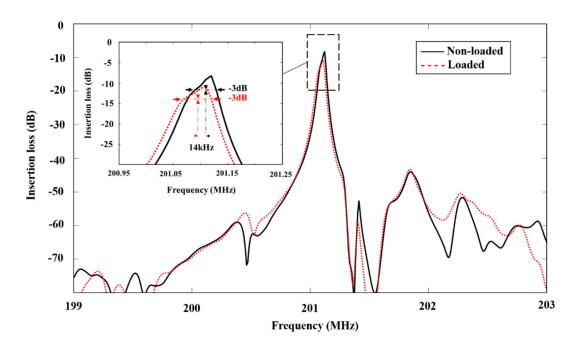


Figure 9. Simulation result of the frequency response before and after loading.

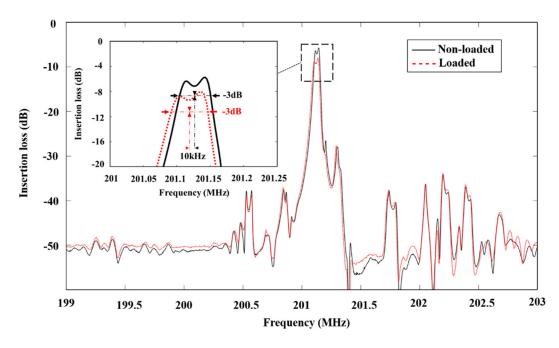


Figure 10. Measured result of the frequency response before and after loading.

Furthermore, there was a larger passband ripple in the measured frequency response. According to the analysis in previous sections, the main peak of TC- SAWR's frequency response was composed of the 0-order symmetric resonant mode ( $S_0$ ) and the 1-order antisymmetric resonant mode ( $A_1$ ), which means that the passband ripple was caused by the large frequency difference between two resonant modes. The deviation of resonant modes' frequencies mainly depends on the width of the central busbar and the aperture size of the TC-SAWR. In order to reduce the frequency difference between the two resonant modes and increase the Q-value of TC-SAWR, optimizations of the structure will be deployed in the future.

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### 4. Conclusions

In the field of biosensors, small sizes of substances to be measured, such as cells and proteins, are common. However, small size loads often cause waveform distortion in longitudinally coupled resonators (LC-SAWR) and other sensitive chips. Further analysis shows that the cause of the waveform distortion is that the small size loads cannot completely cover the sensitive area of LC-SAWR, resulting in uneven load distributions on the surface of the sensor. In order to suppress the waveform distortion caused by the small size load, a transversely coupled SAW resonator (TC-SAWR) was designed and fabricated in this paper. For TC-SAWR, acoustic wave resonates transversely in its aperture, and the mass loading acting on the TC-SAWR will change the resonance frequency of each transverse mode, regardless of whether it covers the aperture completely or not, which avoids waveform distortions caused by the small size loads.

On the other hand, accurate simulation of TC-SAWR has been a difficult problem for a long time. In this paper, 2D-COM theory and the finite element method were used to realize the precise simulation of TC-SAWR. The COM-parameters and the anisotropic parameters were calculated, then the 2D-COM equations of TC-SAWR were established. Finally, by using the PDE module of COMSOL, the 2D-COM equations could be solved numerically, and the response of TC-SAWR under SiO<sub>2</sub> load was obtained.

For comparison and verification, TC-SAWR with a  ${\rm SiO_2}$  cylinder attached to its surface was fabricated and measured. Both the simulation result and the experiment result confirmed the effectiveness of the TC-SAWR device to conquer waveform distortions in the detection of small size load.

This paper illustrated the advantage of TC-SAWR in the field of biochemical detection and introduced an effective method (2D-COM combined with FEM) to simulate the device's performance. Based on this investigation, TC-SAWR is expected to play an important role in biochemical detection.

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