

Review

A Different Approach for Maximum Power Point Tracking (MPPT) Using Impedance Matching through Non-Isolated DC-DC Converters in Solar Photovoltaic Systems

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Abstract: This paper presents a maximum power point tracking (MPPT) algorithm in a solar photovoltaic (SPV) system that requires fewer sensors, is easy to implement, and offers a good tracking efficiency and speed. Among various MPPT algorithms found in literature, the perturb and observe (P&O) algorithm is most popular as it required less complex circuitry and is easy to implement, but it suffers against fast-changing environmental conditions mostly in the case of partial shading conditions and requires two sensors, i.e., voltage and current. This paper presents the one sensor-based method based on the modified converter design which eliminates the effect of partial shading conditions as well as the fast-changing environmental condition. The presented MPPT algorithm was implemented and tested using an FPGA-based NI-sbRIO card interfaced with *Labview* software.

Keywords: impedance matching; maximum power point tracking (MPPT); DC-DC converter; solar PV (SPV) systems



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1. Introduction

In the following decade, energy demand has surged enormously worldwide. This resulted in an arduous challenge of gush in energy demand as a result of overall socio-economic growth [1]. Prominent among prevailing renewables, the photovoltaic cell (solar PV) is on the leading edge [2], but the efficiency of the solar PV structure is not adequate so as to boost the efficiency; it must be driven at a point where we can extort the utmost potential power from it. Current literature reveals that research exertions are being made to boost the output of modules in the context of MPPT. The characteristics of an SPV module are non-linear and its power vs. voltage characteristics indicate that there subsists a solitary point (P_m) at which the module delivers maximum power [3]. The point P_m deviates due to change in temperature and/or insolation [4,5]. Consequently, the mismatch between the source and load characteristics lessens the availability of the utmost potential accessible power transmitted to the load, leading to huge power losses. To minimize power loss, the MPPT algorithm ensures that the impedance of the source will be equal to or near to the load impedance which is done by the DC-DC converter. Classification and comparison of MPPT techniques are also available in the literature [6–8]. MPPT techniques in solar photovoltaic systems under uniform insolation [9–11] and partially shaded conditions [12–14] are also compared in the literature.

1.1. Operating Strategy of MPPT

The functioning of MPPT has explicated an example for tracking maximum power through the changes in insolation, as depicted in Figure 1a. Figure 1b depicts characteristics of a solar cell for a linear resistive-type load ($R_L = 10 \Omega$); it has three maximum power points (MPPs) which are A' , B' , and C' on three different insolation level. Since the load is linear, the operating points of the cell and the respective terminal voltages are A, B, and C. It is apparent from Figure 1a,b that the power delivered via the solar cell concerning A, B, and C is less than the power available [15].

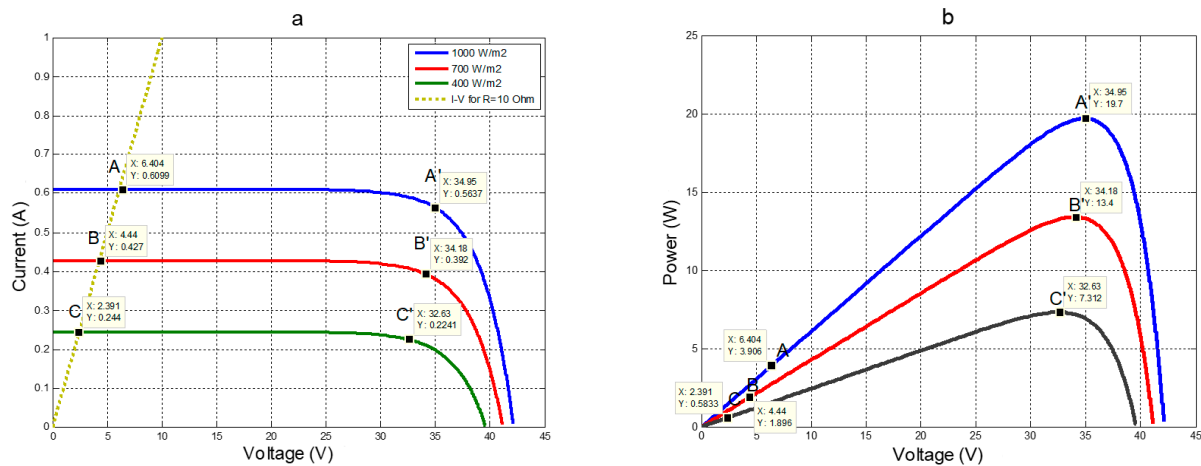


Figure 1. MPP for different insolation on (a) current vs. voltage curve and (b) power vs. voltage curve.

The MPPT functioning evokes the concept of seizing the terminal voltage corresponding to the MPPs which are A' , B' , and C' in place of the operating points A, B, and C. To have the operating point of the solar cell at the MPP, an electronic circuitry recognized as the DC-DC converter was utilized in MPPT [16–18].

1.2. Impedance Matching by DC-DC Converter in MPPT Operation

In the tracking of MPP, the DC-DC converter is a vital component as the source is capable of delivering the utmost power only corresponding to V_{MPP} or I_{MPP} . Although, in designing a DC-DC converter, it is quite problematic to move the terminal voltage (V_i) or current (I_i) of an SPV module in consequence to a V_{MPP} or I_{MPP} . This paper demonstrates the MPPT algorithms based on various designs of the DC-DC converter [15] in which the impedance as perceived from a solar PV (Z_{source}) matches the analogous impedance at the MPP ($Z_{mpp} = V_{mpp}/I_{mpp}$). Figure 2 depicts the block illustration of the general MPPT operation. A comparison of MPPT algorithms and the control of DC-DC converters is presented in [19–22].

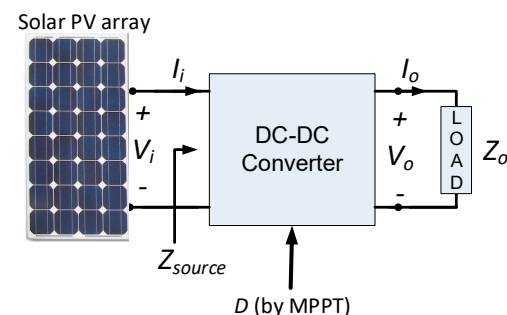


Figure 2. Block illustration of general MPPT operation [15].

To authenticate the response of the converters, simulations are accomplished in MATLAB's SIMSCAPE library, which is closer to the practical design. The N-channel MOSFET is

utilized for the switching function and it is directed through a voltage-controlled PWM generator. The general operation of MPPT is depicted in Figure 2, wherein the input constraints are sensed and processed by the MPPT controller, thus the duty cycle is set accordingly by the MPPT algorithm. The DC-DC converter works at this duty cycle to deliver the utmost power to the load. More than 31 MPPT techniques have been found in the literature and of these 31 techniques, the perturbation and observation (P&O) or hill climbing is well-liked, with the benefits of ease of realization and involving less complex circuitry.

1.3. P&O/Hill Climbing Technique of MPPT

The P&O technique is one of the well-liked MPPT techniques. The technique is fundamentally an iterative approach. In this technique, the operating point of the SPV module swings around the MPP. The power vs. voltage curve of a solar PV shows that the change in power in regards to voltage (dP/dV) is positive, negative, and zero for the province before the MPP, after the MPP, and at the MPP, respectively [23,24].

This technique is realized by perturbing the duty ratio at regular intervals and swinging around the point $dP/dV = 0$, i.e., MPP. This operation is explicated in Figure 3 and the methodology is given in Table 1.

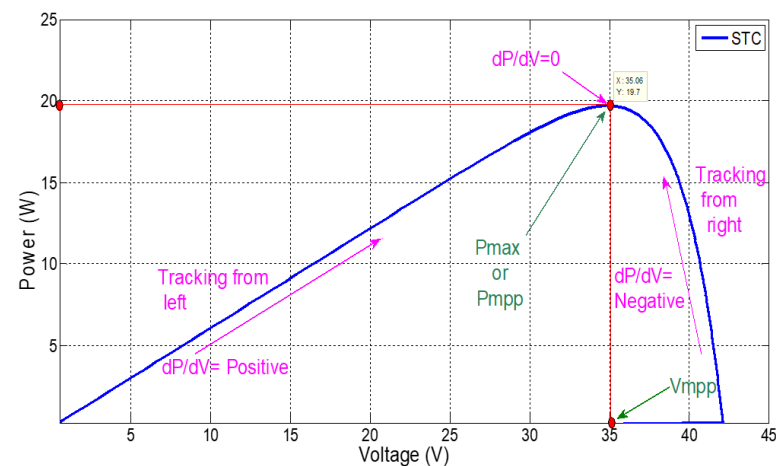


Figure 3. Power characteristic of SPV in perspective with P&O MPPT [23].

Table 1. Methodology of P&O technique [23].

Perturbation	Change in Power	Next Perturbation
(+) Positive	(+) Positive	(+) Increment in duty ratio 'D'
(+) Positive	(−) Negative	(−) Decrease in duty ratio 'D'
(−) Negative	(+) Positive	(−) Decrease in duty ratio 'D'
(−) Negative	(−) Negative	(+) Increment in duty ratio 'D'

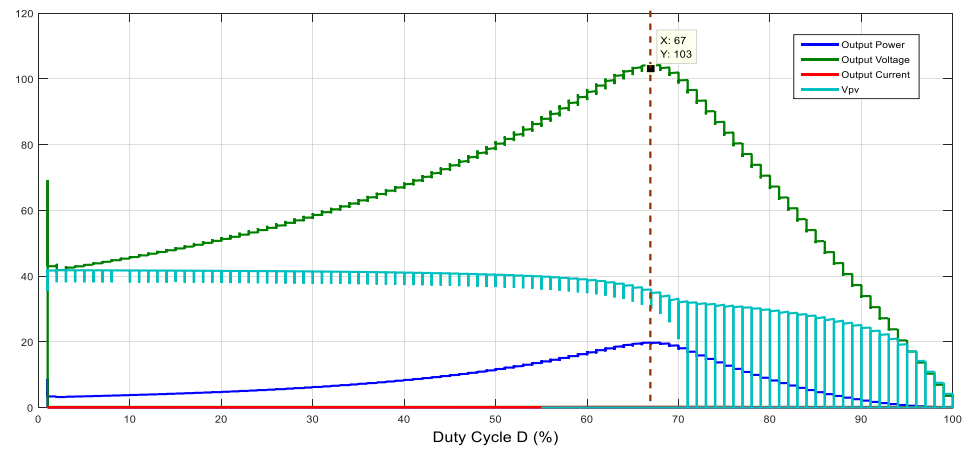
2. Proposed MPPT Technique

The proposed MPPT technique is based on a modified converter design [15] as the DC-DC converters are playing a very important role in MPPT. The design of DC-DC converters is an essential part of any MPPT scheme [25,26]. Application of conventional DC-DC buck converters [27,28], boost converters [29–31], buck-boost converters [32–34], novel converters [35,36], and multilevel inverters [37,38] are also reported in the literature for MPPT operation.

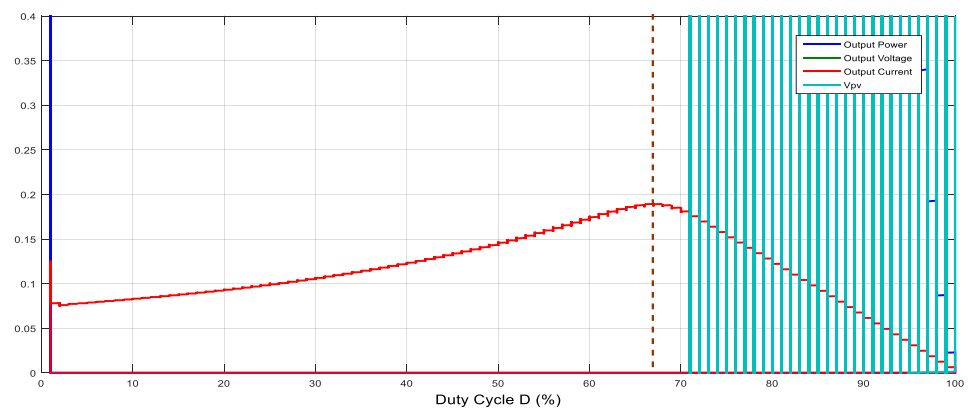
The behavior of DC-DC converters is observed for variation in the duty cycle for the buck, boost, and buck-boost converters in this paper and is explained in the subsequent section.

2.1. Boost Converter: Relation between Output Voltage, Output Current, Output Power, and V_{PV}

A graph of the output voltage, output current, output power, and V_{PV} with variation in the duty cycle is presented in Figure 4. It shows that the point of maximum power occurs when the load voltage is at the maximum.



(a)



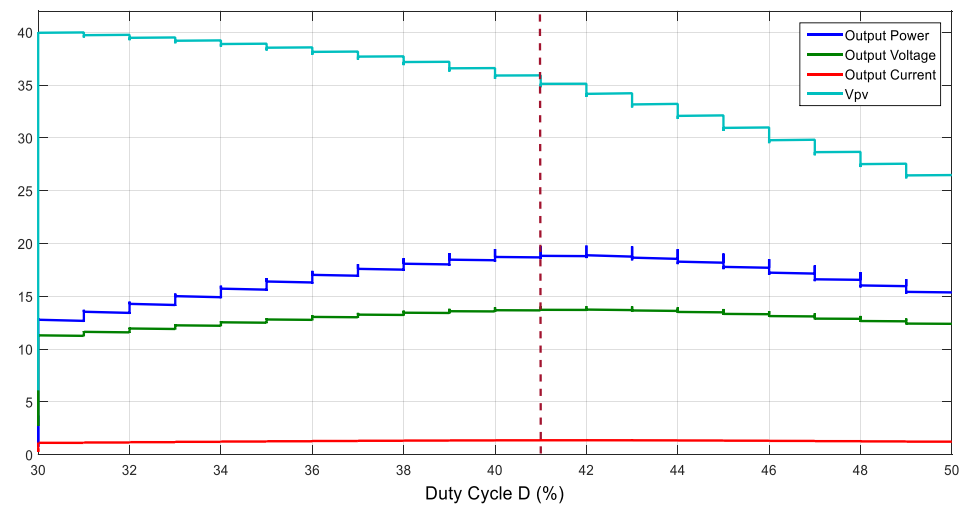
(b)

Figure 4. Boost converter: graph of output current, output power, and V_{PV} with duty cycle.

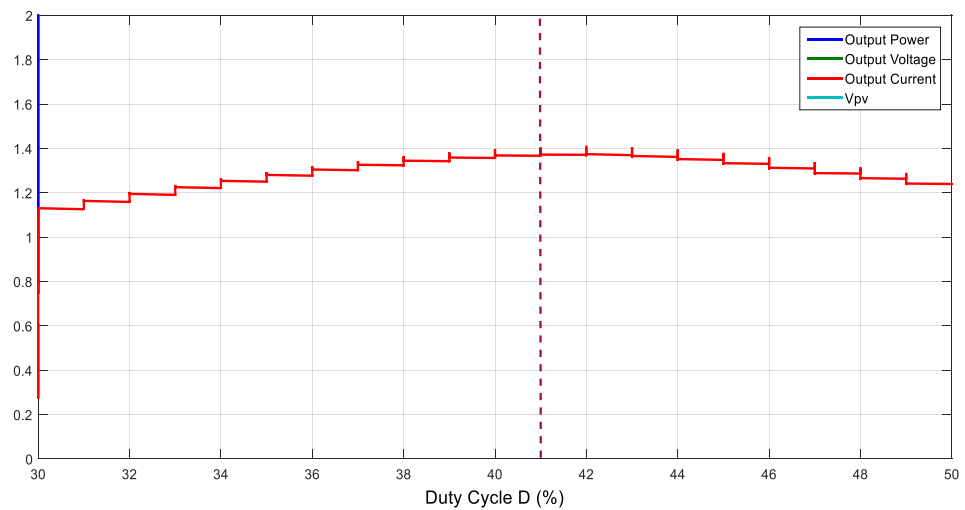
The current is also at the maximum where the power is maximum at $D = 67\%$, which is not visible in Figure 4a; by zooming in at the Y-axis, it is clear in Figure 4b.

2.2. Buck Converter: Relation between Output Voltage, Output Current, Output Power, and V_{PV}

A graph of the output voltage, output current, output power, and V_{PV} with the duty cycle is presented in Figure 5.



(a)



(b)

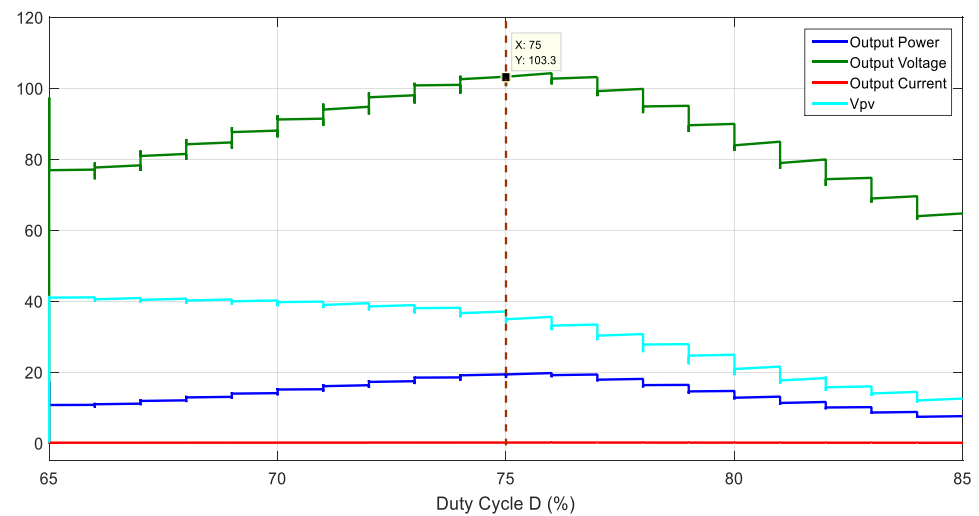
Figure 5. Buck converter: graph of output voltage, output current, output power, and V_{PV} with duty cycle.

Figure 5 shows that the point of maximum power occurs when the load voltage is at the maximum at duty cycle $D = 41\%$ [15].

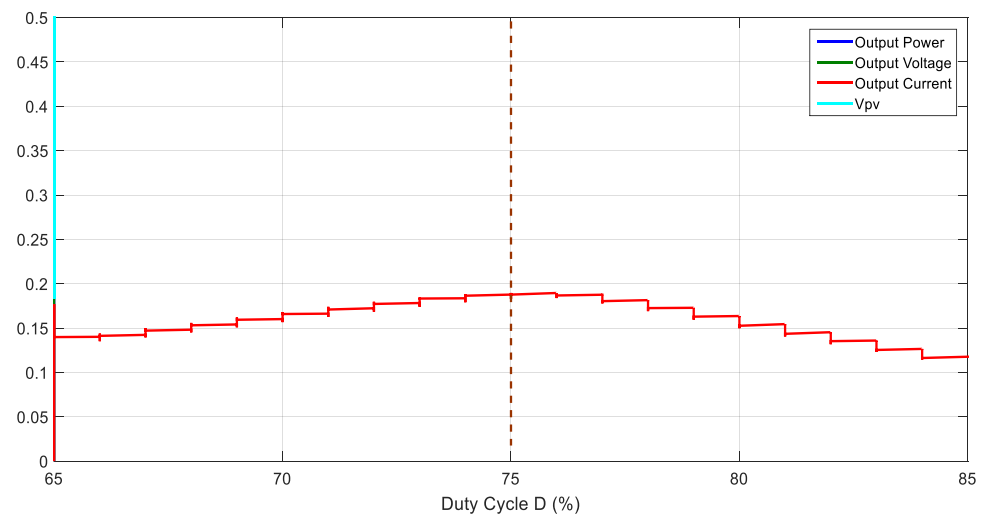
The current is also at the maximum where the power is maximum at $D = 41\%$, which is not visible in Figure 5a; by zooming in at the Y-axis, it is clear in Figure 5b.

2.3. Buck-Boost Converter: Graph of Output Voltage, Output Current, Output Power, and V_{PV} with the Duty Cycle

A graph of the output voltage, output current, output power, and V_{PV} with the duty cycle is presented in Figure 6. It shows that the point of maximum power occurs when the load voltage is at the maximum at duty cycle $D = 75\%$ [15]. The current is also at the maximum where the power is at the maximum at $D = 75\%$, which is not visible in Figure 6a; by zooming in at the Y-axis, it is clear in Figure 6b.



(a)



(b)

Figure 6. Buck-boost converter: graph of output voltage, output current, output power, and V_{PV} with duty cycle.

2.4. Operating Principle of Proposed MPPT Technique

The conventional P&O method senses input parameters or PV side parameters, i.e., V_{PV} and I_{PV} , and compares power at different time intervals as well as keeps on oscillating around the maximum power. The maximum power will be transferred to the load when the load side voltage will be at the maximum. An MPPT technique is proposed which maximizes the load voltage in contrast to V_{PV} in conventional P&O and causes maximum power to be delivered to the load, as shown in Figure 7. The calculated duty cycle for standard environmental conditions gives the maximum power at which the load voltage is at the maximum; this duty cycle becomes the maximum duty cycle (D_{max}). The minimum duty cycle (D_{min}) will be calculated by assuming the minimum set of environmental conditions, such as, for example, an insolation of 50 W/m^2 and temperature of 25°C . Calculation of D_{min} and D_{max} depends on the location of the solar PV array. In the proposed MPPT technique, the perturbation of the duty cycle is done within the range of D_{min} to D_{max} to maximize the load voltage; in this way, the proposed method becomes the PV array-dependent method. Nevertheless, the response time to achieve MPP becomes less.

This novel method is applied by perturbing the duty ratio at regular intervals and observing the load voltage, thus oscillating around the point $V_o \cong V_{o(Max)}$, i.e., MPP. The operation is explained in Figure 7 and Table 2.

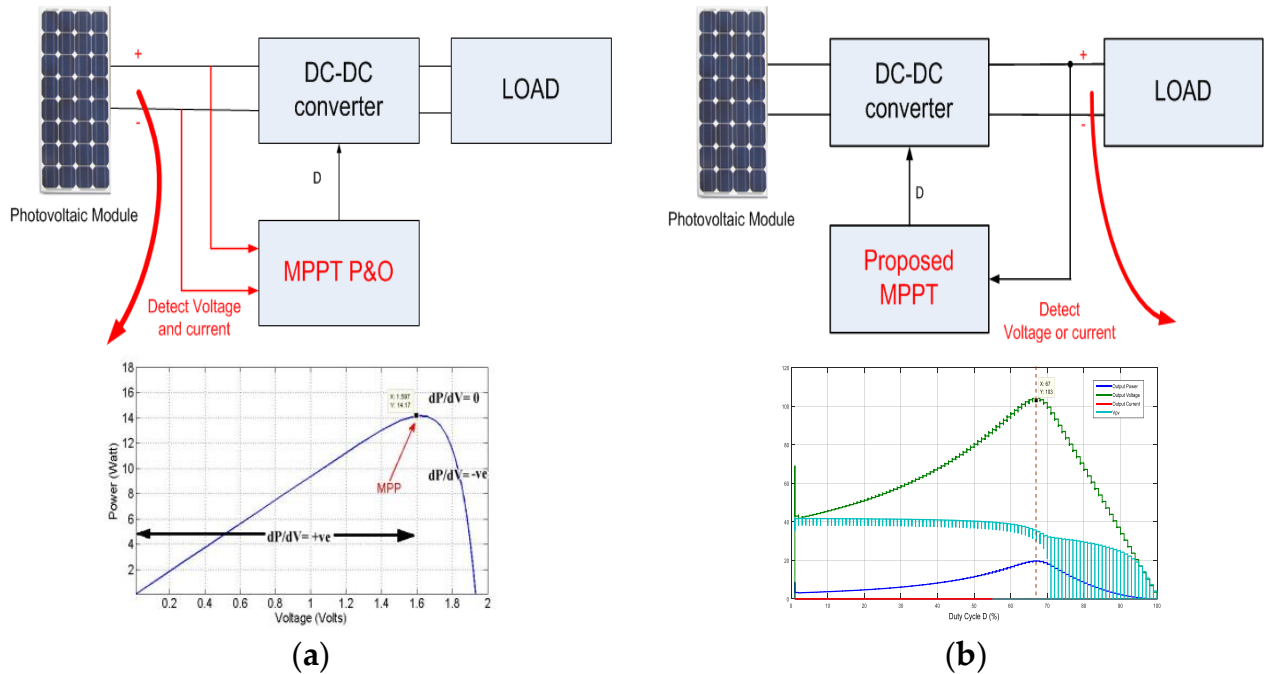


Figure 7. Comparison of (a) conventional P&O method with (b) proposed method in terms of sensing parameters.

Table 2. Methodology of the proposed method.

Perturbation	Change in Load Voltage	Next Perturbation
(+) Positive	(+) Positive	(+) Increment in duty ratio 'D'
(+) Positive	(−) Negative	(−) Decrease in duty ratio 'D'
(−) Negative	(+) Positive	(−) Decrease in duty ratio 'D'
(−) Negative	(−) Negative	(+) Increment in duty ratio 'D'

2.5. Flow Chart of Proposed MPPT Algorithm

Figure 8 shows the flow chart of the proposed MPPT method. The method applies the perturbation in the duty cycle within a predetermined range (D_{min} to D_{max}) to regulate the output voltage of the converter or the load side voltage. The initial conditions are assumed to be zero, which means the previous value of the voltage and duty cycle is zero. The algorithm starts with the minimum value of the duty cycle (D_{min}); senses the load voltage, which is V_{new} ; and compares it to the previous value of the load voltage (V_{old}) if $V_{new} > V_{old}$ and $D_{new} > D_{old}$. Then, the next perturbation is the positive increment of the duty cycle. The MPPT operation follows the methodology presented in Table 2.

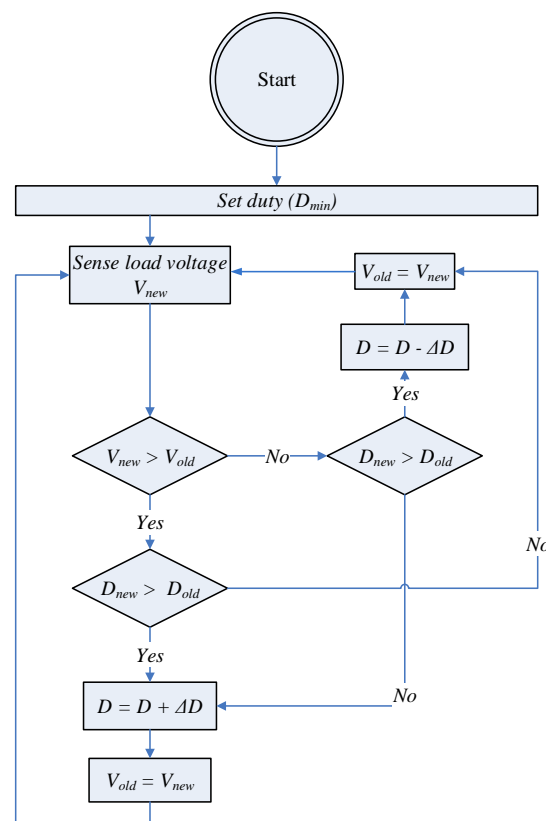


Figure 8. Flow chart of proposed MPPT algorithm.

3. Hardware Implementation of Proposed MPPT Method

The proposed scheme of MPPT is developed using a Field-Programmable Gate Array (FPGA)-based NI-sbRIO card interfaced with *Labview* software. Control signals were generated using *Labview*, which operates the boost converter to track the maximum power, as shown in Figure 9. The proposed algorithm was tested using a boost converter because the converters with continuous input current (as in the case of the boost converter) are preferable from the PV array point of view.

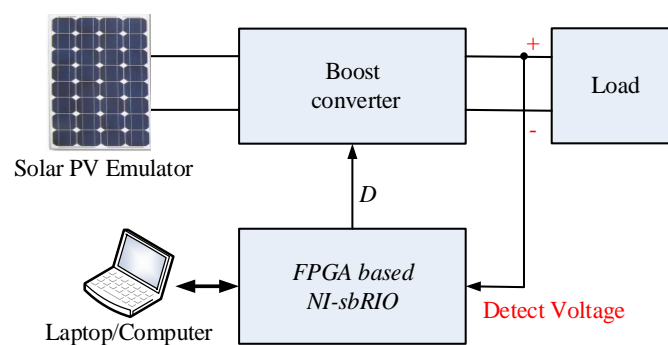


Figure 9. Block diagram of hardware scheme.

Figure 9 depicts the block diagram of the proposed MPPT algorithm, in which sensing of the load voltage is done to track MPP as compared to the convention P&O method, which senses the voltage and current of the source-side or PV array side.

3.1. Solar PV Emulator

The solar PV array is emulated by employing a solar PV emulator made by *igrenEnergi* and *ecosense world* [39,40], as shown in Figure 10.



Figure 10. Solar PV emulator with four channels (CH1 to CH4).

The solar PV emulator is a programmable and controlled power supply designed to emulate solar panels. The PV emulator maintains the output corresponding to the I - V characteristic of the panel with a fast transient response to change in load conditions for a given environmental condition. It emulates the output of solar panels from different manufacturers, variations due to the time of the day, the effect of season, and different geographical locations. The PV emulator gives output across its four channels to match the solar panel data selected by the user based on the date and time.

The specification of the array taken for testing is given in Figure 11a, wherein panel (SR-M572175 made by Sunrise Solartech Co. Ltd., Changzhou 21302, China) details including standard temperature and insolation are $I_{MPP} = 4.98$ A and $V_{MPP} = 35.2$ V, as shown in Figure 11b. The day selected (10 July 2015) and time (9:00 A.M.) give the operating data $I_{MPP} = 1.754$ A and $V_{MPP} = 32.4$ V at insolation 352 W/m² and temperature 26.05 °C, and the simulation (I - V and P - V characteristic) obtained from the emulator is shown in Figure 12.

Simulation details input

Select device
71

Select channel
0

Select manufacturer
Sunrise Solartech Co. Ltd.

Select panel
SR-M572175

Simulation month and day
Month 7 Day 10

Start time
8 0

End time
14 0

Latitude
23.2599

Longitude
77.4126

Panel tilted angle
30

Type of surrounding (Albedo Factor)
0

Generate simulation

(a)

Panel details

Imp	Vmpp	ISC	VOC	Alpha	Beta	Temperature	Fill factor	Global radiation	Diffuse radiation
4.98	35.2	5.43	43.3	0.002715	-0.14722	25.81	0.746	7301.579	1464.737

Simulation IV table

8 9 10 11 12 13

Import **Export**

Selected hour	Temperature	Insolation
9	26.05	352

(b)

Figure 11. (a). Specification of the array taken for testing. (b). Specification of the array uploaded in the PV emulator.

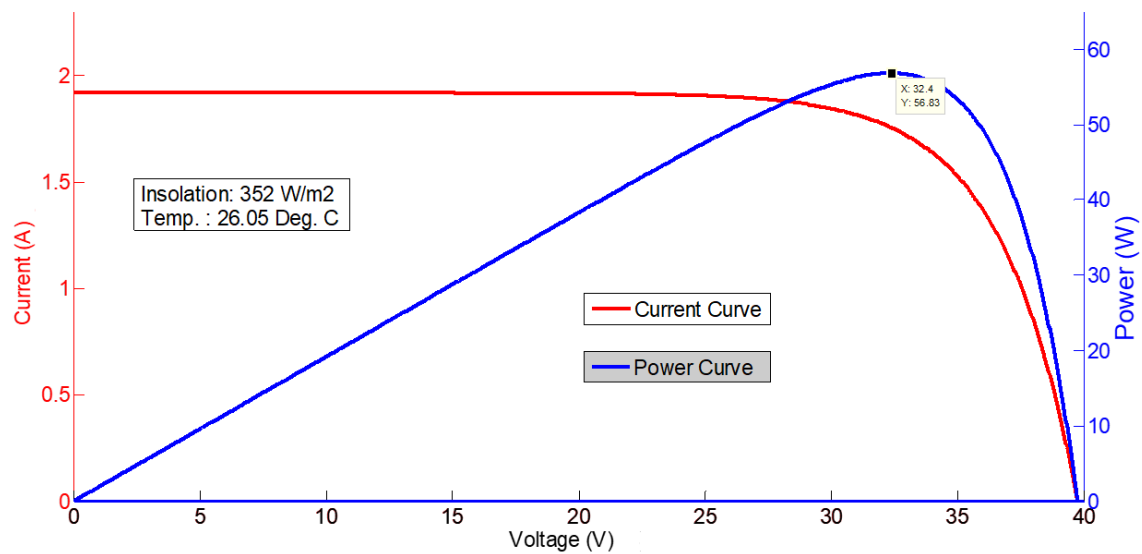


Figure 12. *I-V* and *P-V* characteristic of PV array at 09:00 A.M.

The PV array emulation was used for a different time for testing purposes across the emulator's different channels, as shown in Figure 13. In channel 0 (CH0), the *I-V* characteristic table or *I-V* table was downloaded for time 09:00 A.M.; in channel 1 (CH1) for time 10:00 A.M.; in channel 2 (CH2) for time 11 A.M.; and in channel 3 (CH3) for time 08 A.M.

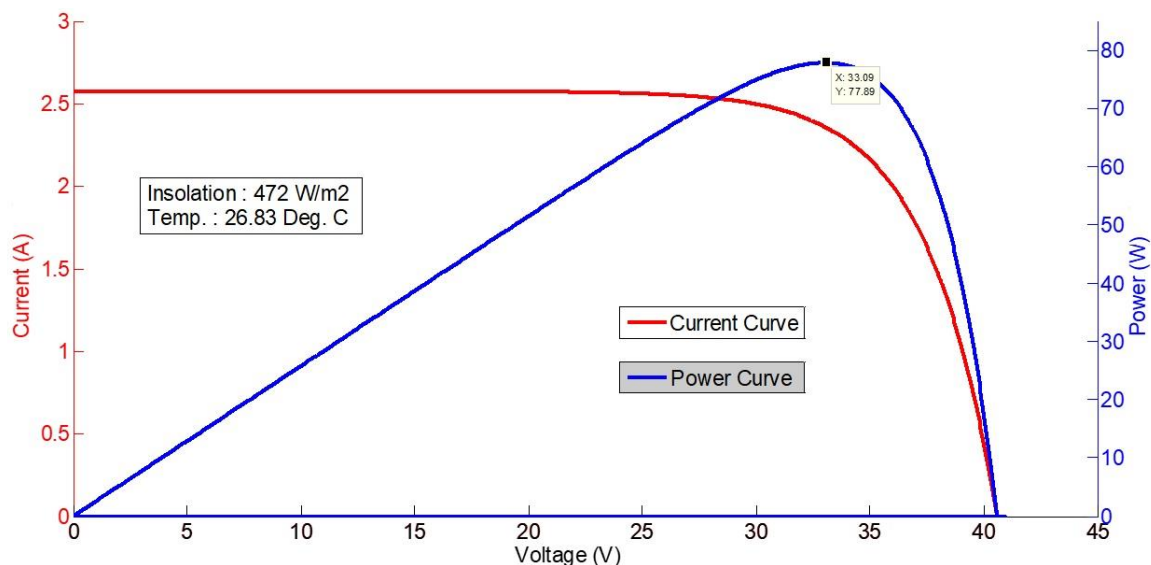


Figure 13. *I-V* and *P-V* characteristic of PV array at 10:00 A.M.

For the time 10:00 A.M., the emulator gave the operating data $I_{MPP} = 2.354$ A and $V_{MPP} = 33.09$ V at insolation 472 W/m^2 and temperature 26.83°C , and the simulation (*I-V* and *P-V* characteristic) obtained from the emulator is shown in Figure 13. The *I-V* table was uploaded across channel 1.

3.2. DC-DC Boost Converter

The DC-DC converter design was presented in [15] and the duty cycle of the boost converter was calculated for the emulated PV array at time 13:00. For time 13:00, the operating data obtained was $I_{MPP} = 2.904$ A and $V_{MPP} = 34.9$ V, and $Z_{MPP} = 12.04$ at the insolation level of 648 W/m^2 and temperature 27.11°C . The *I-V* table uploaded at channel 3 is shown in Figure 14.

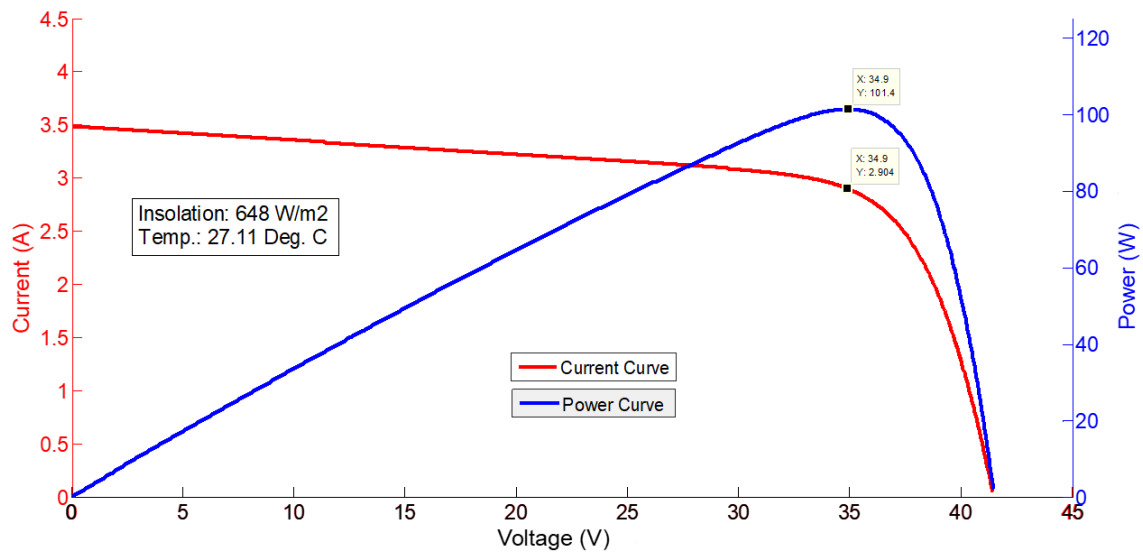


Figure 14. *I-V* and *P-V* characteristic of PV array at 13:00.

The duty cycle can be calculated as:

$$D = 1 - \sqrt{\frac{Z_{MPP}}{Z_o}} = 1 - \sqrt{\frac{12.04}{40}} = 0.4514$$

The output voltage or load voltage can be calculated as

$$V_o = \frac{V_i}{(1 - D)} = 63 \text{ V}$$

and maximum power is calculated as

$$P_{max} = V_{MPP} \times I_{MPP} = 34.9 \times 2.904 = 101.4 \text{ W}$$

The DC-DC boost converter used for the testing is shown in Figure 15. The output of the PV array obtained from the PV emulator was sent to the DC-DC boost converter and controlling of the converter through the duty cycle of the boost converter was done by an FPGA-based *NI-sbRIO* card.

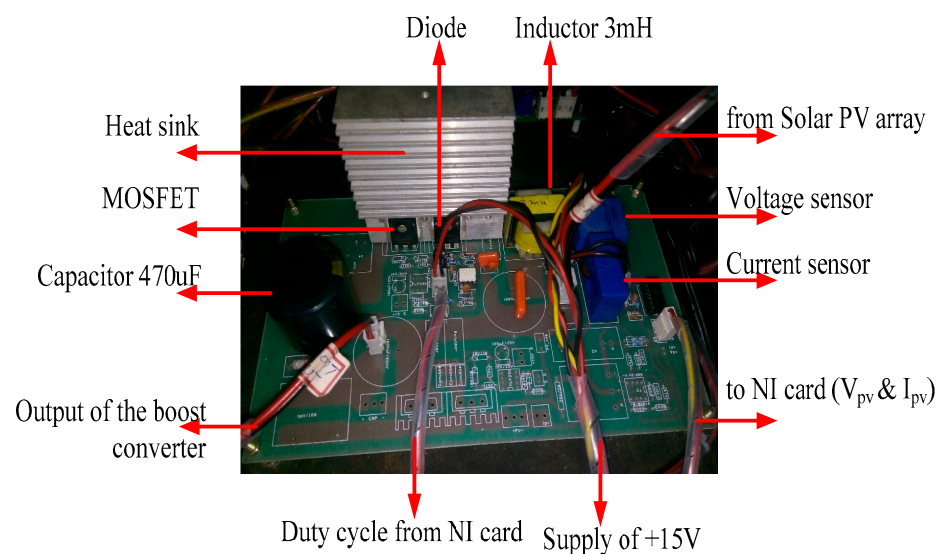


Figure 15. DC-DC boost converter.

3.3. MPPT Operation by FPGA-Based NI-sbRIO Card Interfaced with LABVIEW

The *sbRIO-9606* embedded device [41] was used to perform the MPPT operation of the proposed method. The *sbRIO-9606* is an embedded acquisition and control device which integrates a user-reconfigurable FPGA and input–output (I/O) on a single-printed circuit board (PCB) [41]. The connection diagram of the *sbRIO-9606* [41] card is shown in Figure 16.

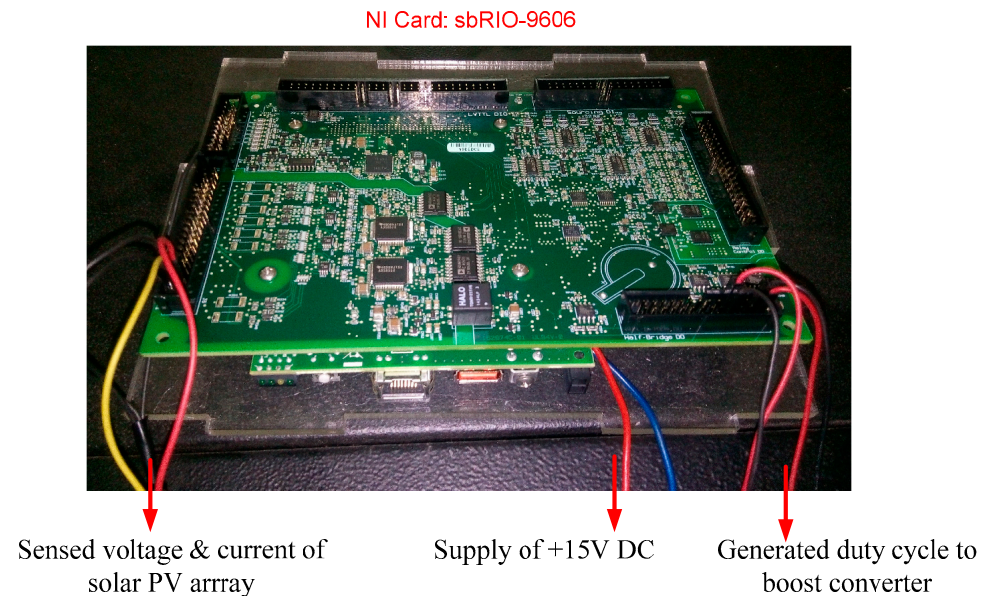


Figure 16. NI card: *sbRIO-9606* connection diagram.

The *Labview* implementation of the proposed MPPT control is shown in Figure 17. Figure 17 illustrates the *Labview* schematic used to implement the MPPT algorithm. The MPPT control was achieved by modulating the duty ratio of the DC-DC converter.

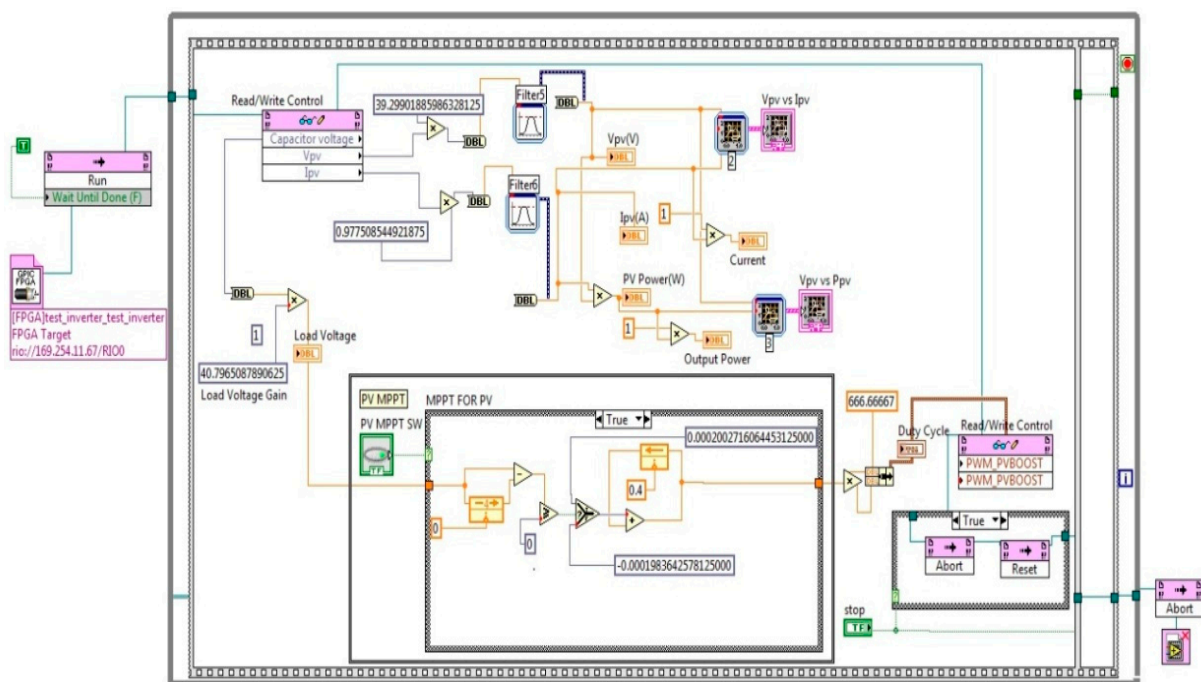


Figure 17. *Labview* block diagram of proposed MPPT technique.

3.4. Resistive-Type Load

The linear resistive-type load was considered for testing. Three 600 W, 100 Ω variable rheostats were used in the experimentation. The value of 40 Ω resistance was obtained by these rheostats.

Figure 18 depicts the complete setup of the hardware implementation; the control circuit or MPPT algorithm was developed in *Labview* software; and to operate the boost converter, the duty cycle was generated by the NI card. A computer was interfaced with the PV emulator through the USB cable and interfacing with the NI card was done through the LAN cable. Results were analyzed through the YOKOGAWA digital signal oscilloscope DL750, whereas the readings were taken through a multi-meter and ammeter.

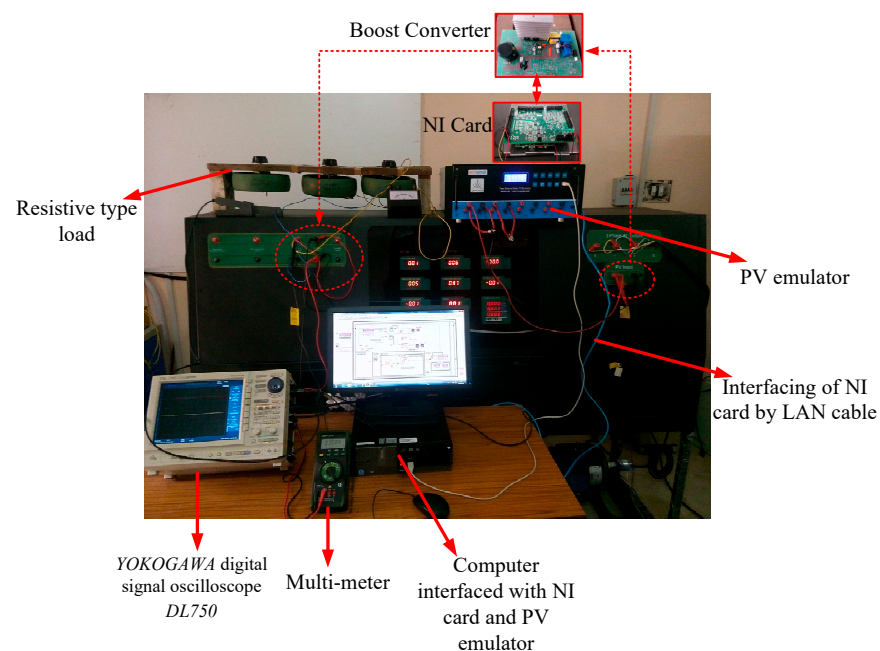


Figure 18. Complete setup of hardware implementation.

4. Results and Discussion

Figure 19 shows the front window of the *Labview* software, whereas Figure 17 shows the back window of the *Labview*.

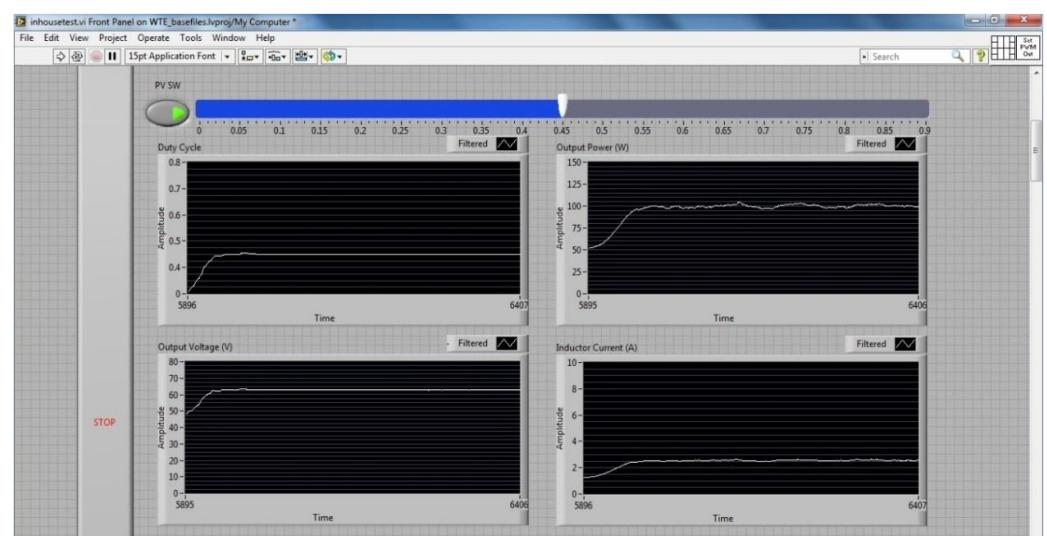


Figure 19. The front window of *Labview*.

The left-hand side top window exhibits the duty cycle modulation and becomes stable at $D = 0.45$ when its operating point approaches MPP, corresponding to a solar isolation of 648 W/m^2 (13:00). The right-hand side top window shows the measured PV array output power across the load. The lower LHS window shows the PV output voltage as 62.5 V , whereas the RHS window displays the inductor current of the boost converter in continuous conduction mode.

The results across the load resistance of 40Ω were taken on the YOKOGAWA digital signal oscilloscope DL750, as shown in Figure 20.

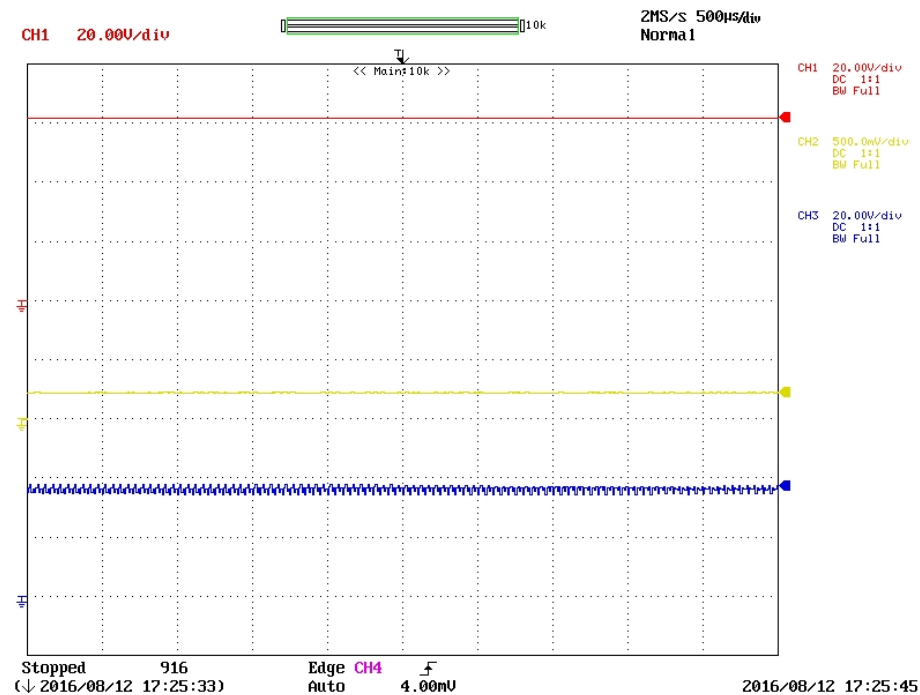


Figure 20. Results (for 13:00) obtained by hardware implementation: CH1: voltage across the load resistance (62.5 V); CH2: current across the load (1.6 A); and CH3: solar PV array's terminal voltage ($V_{PV} = 34.9 \text{ V}$).

Channel 1 (CH1) shows the voltage across the load resistance, channel 2 (CH2) shows the current across the load, and channel 3 (CH3) shows the PV array's terminal voltage (V_{PV}), and the results show that the proposed MPPT technique tracks the maximum power of 100 W .

The selected day (10 July 2015) and time (9:00 A.M.) give the operating data $I_{MPP} = 1.754 \text{ A}$ and $V_{MPP} = 32.4 \text{ V}$ at insolation 352 W/m^2 and temperature $26.05 \text{ }^\circ\text{C}$. Figure 12 gives the I - V and P - V characteristic of the array that indicates that the maximum power is 56.83 W . The proposed algorithm tracks the maximum power, which is shown in Figure 21.

The selected day (10 July 2015) and time (10:00 A.M.) gives the maximum power of 77.89 W , as shown in Figure 13. The proposed algorithm tracks the maximum power and the hardware results are shown in Figure 22.

The proposed MPPT method was further tested for the partial shading condition, wherein two channels of the PV emulator were connected in series (CH1 and CH2), as shown in Figure 23. Two power diodes were used to make the diode mitigation technique and channel 1 gave the PV module characteristic for 10:00 A.M. while channel 2 gave the 11:00 A.M. characteristic, as presented in Figure 23.

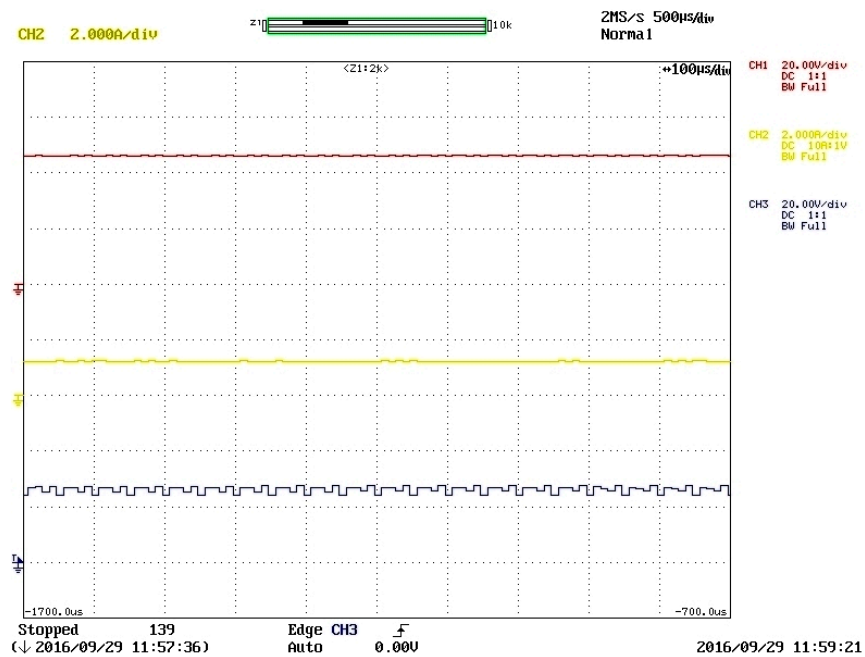


Figure 21. Results (for 09:00 A.M.) obtained by hardware implementation: CH1: voltage across the load resistance (44.8 V); CH2: current across the load (1.25 A); and CH3: solar PV array's terminal voltage ($V_{PV} = 31.5$ V).

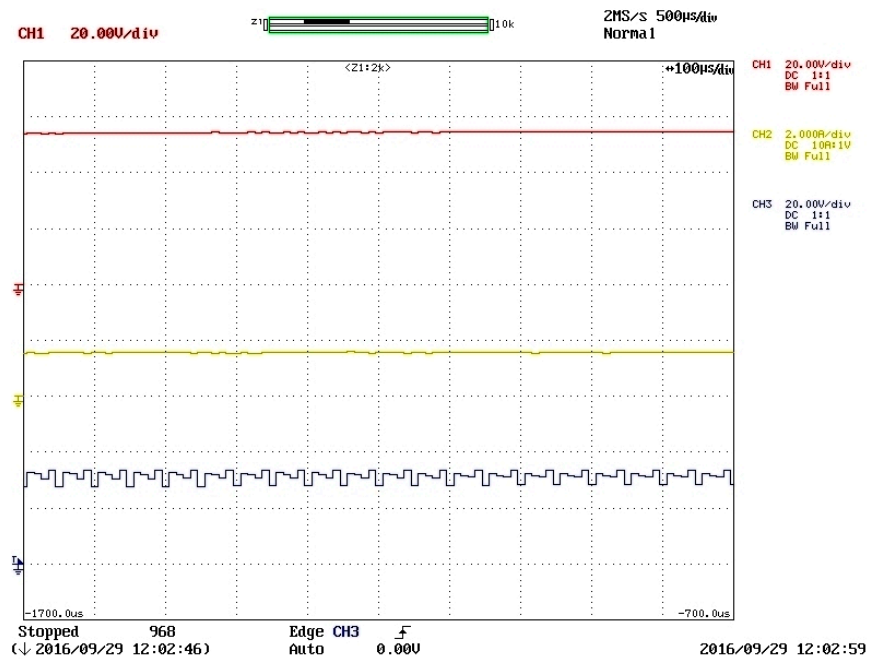


Figure 22. Results (for 10:00 A.M.) obtained by hardware implementation: CH1: voltage across the load resistance (54.5 V); CH2: current across the load (1.4 A); and CH3: solar PV array's terminal voltage ($V_{PV} = 33$ V).



Power diodes

Figure 23. PV emulator connection for partial shading condition.

In this partial shading condition, the I - V and P - V characteristic obtained is shown in Figure 24, whereas the result obtained is shown in Figure 25, and it is clear in the figure that the maximum power achieved from the array was 129 W. However, the most popular MPPT technique of P&O may stick to the local maximum point of 89.46 W and failed to track MPP.

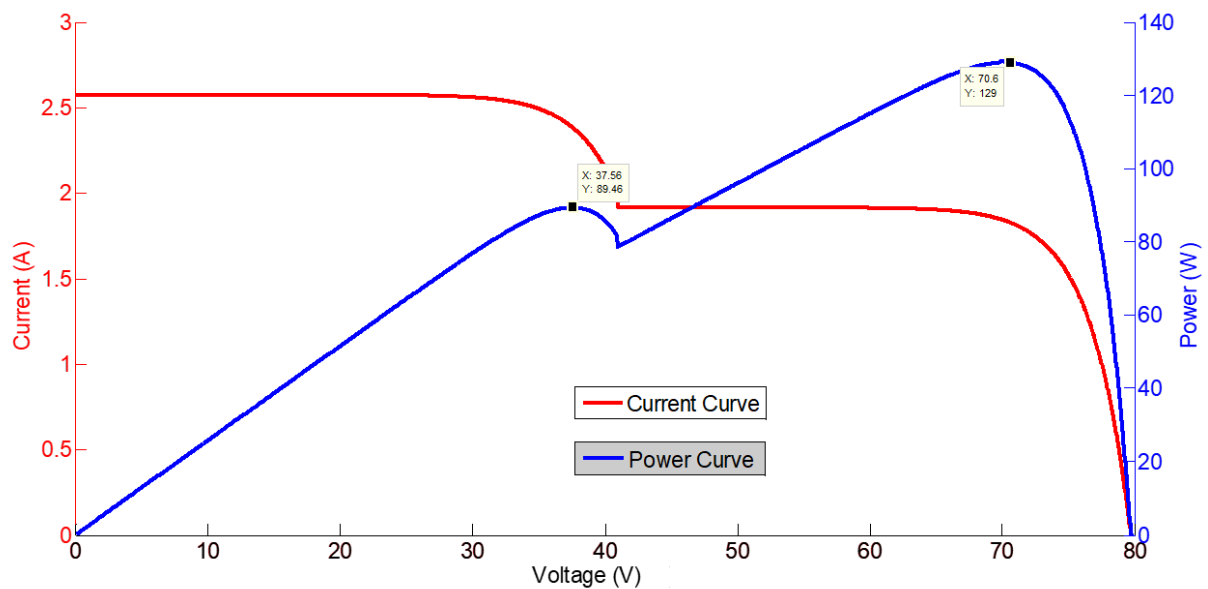


Figure 24. I - V and P - V characteristic of PV array for partial shading condition (CH1 at 10:00 A.M. and CH2 at 11:00 A.M.).

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