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# Adjusting the Single-Diode Model Parameters of a Photovoltaic Module with Irradiance and Temperature

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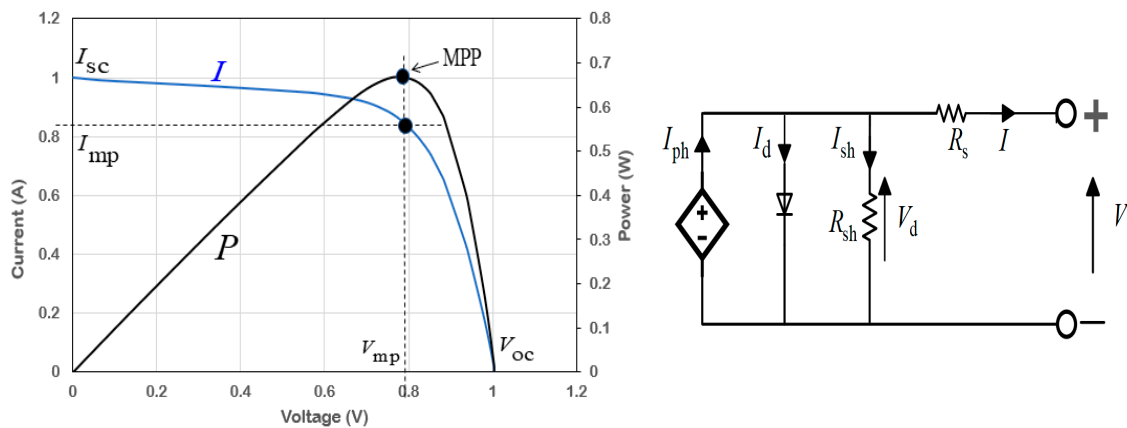
**Abstract:** This paper presents a concise discussion and an investigation of the most literature-reported methods for modifying the lumped-circuit parameters of the single-diode model (SDM) of a photovoltaic (PV) module, to suit the prevailing climatic conditions of irradiance and temperature. These parameters provide the designer of a PV system with an essential design and simulation tool to maximize the efficiency of the system. The parameter modification methods were tested using three commercially available PV modules of different PV technologies, namely monocrystalline, multicrystalline, and thin film types. The SDM parameters of the three test modules were extracted under standard test conditions (STC) using a well-established numerical technique. Using these STC parameters as reference values, the parameter adjustment methods were subsequently deployed to calculate the modified parameters of the SDM under various operating conditions of temperature and irradiance using MATLAB-based software. The accuracy and effectiveness of these methods were evaluated by a comparison between the calculated and measured values of the modified parameters.

**Keywords:** ideality factor; irradiance; photocurrent; saturation current; series resistance; shunt resistance; single-diode model; temperature effects

## 1. Introduction

Photovoltaic systems offer the most direct conversion of the electromagnetic energy in the sunlight into electricity. At the heart of any PV power plant is the PV generator, which typically consists of an array of PV modules connected in series-parallel combinations to deliver the rated power at the required levels of terminal current and voltage. Each PV module consists of several PV cells, which are almost always connected in series to provide a specific terminal voltage. The terminal I-V (current-voltage) and P-V (power-voltage) characteristics of a typical PV generator, whether it is a cell, a module, or an array, are shown in Figure 1. Three salient points can be identified on the I-V curve: The short circuit (SC), the open circuit (OC), and the maximum power point (MPP).

The design of a PV power system requires a lumped-circuit parameters' model of the PV generator. This is used for purposes, such as to properly size the power switching devices used in the power processing converters, designing the maximum power point tracking system, and for efficient sizing of the PV array. In addition, such an equivalent circuit model can be readily embedded in circuit simulation programs. The widely accepted circuit model of a PV generator is the single-diode model (SDM), which is also included in Figure 1 [1–3]. This model can be easily adapted to model a cell, a module, or an array and offers a compromise between accuracy and complexity [4–6]. An extensive study on the measurement uncertainty using different lumped-circuit parameters models of a PV module can be found in [7]. However, this paper was based on the single-diode model.



**Figure 1.** The normalized I-V and P-V curves of a typical PV module (left) and the single-diode model (right).

The SDM involves five parameters: The photocurrent  $I_{ph}$ , the saturation current  $I_{sat}$ , and the ideality factor  $n$  of the diode, the parallel resistance  $R_{sh}$ , and the series resistance  $R_s$ . These parameters are not provided in manufacturers' datasheets of PV modules. However, they can be extracted from data given in a datasheet, which almost always includes the three salient points (SC, OC, and MPP), the number of series-connected cells  $N_s$ , and the temperature coefficients of the short-circuit current  $\mu_{I_{sc}}$  (A/°C) and open-circuit voltage  $\mu_{V_{oc}}$  (V/°C) [8]. However, these parameters are always specified at only one operating condition, namely the STC (STC: Irradiance  $G = 1000$  (W.m<sup>-2</sup>), air mass ratio  $AM = 1.5$ , and temperature  $T = 25$  °C). Since the characteristics of a PV generator can vary significantly with the ambient temperature and irradiance, parameters estimated from information in a datasheet are only valid at standard test conditions. For any other arbitrary values of temperature and irradiance, these parameters must be adjusted, i.e., re-estimated, accordingly [9,10]. It is worth noting at this point that the problem of variations in module parameters is further complicated due to the effects of partial shading (PS) [11,12], which can have adverse consequences on the performance and energy yield of a PV plant. Alleviation of the effects of PS requires the use of bypass diodes. However, these diodes can vary the parameters of the PV generator and also give rise to complicated P-V curves with multiple power peaks [13,14]. However, the focus of this paper was on exploring the performance of methods for varying the parameters of the SDM with variations in irradiance and temperature without including the effects of bypass diodes.

Two different approaches have been reported in the literature for considering the variations of the SDM parameters with fluctuations in ambient climatic conditions [15]. The first approach assumes that the photocurrent depends only on irradiance, while the saturation current depends only on temperature, while all other parameters remain constant [16,17]. The second approach assumes that the saturation current and the resistances vary with both conditions of irradiance and temperature [18]. The ideality factor is, however, assumed to remain constant in both approaches. The contributions of this paper are: Review the major methods of modifying the parameters of the SDM with due account taken of their variations with temperature and irradiance, provide the mathematical derivation of these methods, which is not generally provided in the literature, apply these methods to PV modules of different technologies, and compare their results.

Following this introduction, Section 2 presents the mathematical background that underpins the rest of the work discussed in this paper. The most reported methods of modifying the STC parameters of the SDM according to changes in irradiance and temperature, referred to henceforth as the parameter adjustment methods, are presented and discussed in Section 3. Section 4 presents the outcome of synthesizing the results of these adjustment methods when applied to three PV modules of different PV technologies and compares these results with measured values. Finally, Section 5 summarizes the outcomes of this article.

## 2. Mathematical Analysis of the Single-Diode Model

This section presents the mathematical analyses of the single-diode model that pertains to the parameter adjustment methods, which are the subject of the next section.

A PV module with  $N_s$  series-connected PV cells can be modelled using the SDM depicted in Figure 1. The terminal current  $I$  (A) and voltage  $V$  (V) of the module are expressed by the nonlinear, implicit, and transcendental equation:

$$I = I_{ph} - I_{sat} \left[ \exp\left(\frac{V + IR_s}{nN_s V_{th}}\right) - 1 \right] - \frac{V + IR_s}{R_{sh}} \quad (1)$$

Referring to Figure 1, the current  $I_{ph}$  represents the photocurrent, which is mainly dependent on the incident irradiance  $G$  [19]. The second term is the diode current  $I_d$ , which is represented by Shockley's equation, and the last term is the current  $I_{sh}$ , which flows in the parallel resistance  $R_{sh}$ . Finally,  $V_{th}(V)$  is the thermal equivalent voltage, which is given in terms of the electronic charge  $q = 1.602 \times 10^{-19}$  C, the temperature  $T$ (K), and Boltzmann's constant  $k = 1.3806503 \times 10^{-23}$  J/K as:

$$V_{th} = kT/q. \quad (2)$$

Substituting  $(V_{oc}, 0)$  in Equation (1), we obtain the open-circuit equation:

$$\text{OC : } I_{ph} = I_{sat} \left[ \exp\left(\frac{V_{oc}}{nN_s V_{th}}\right) - 1 \right] + \frac{V_{oc}}{R_{sh}}. \quad (3)$$

Since  $\frac{V_{oc}}{nN_s V_{th}} \gg 1$ , the  $(-1)$  in the exponential term can be neglected [20]. Hence, re-arranging, we can express the open-circuit voltage as:

$$V_{oc} = nN_s V_{th} \ln\left(\frac{I_{ph} R_{sh} - V_{oc}}{I_{sat} R_{sh}}\right). \quad (4)$$

Substituting  $(0, I_{sc})$  in Equation (1), we obtain the short-circuit equation:

$$\text{SC : } I_{sc} = I_{ph} - I_{sat} \left[ \exp\left(\frac{I_{sc} R_s}{nN_s V_{th}}\right) - 1 \right] - \frac{I_{sc} R_s}{R_{sh}}. \quad (5)$$

Finally, substituting the MPP point  $(V_{mp}, I_{mp})$  into Equation (1), we obtain the MPP equation:

$$\text{MPP : } I_{mp} = I_{ph} - I_{sat} \left[ \exp\left(\frac{V_{mp} + I_{mp} R_s}{nN_s V_{th}}\right) - 1 \right] - \frac{V_{mp} + I_{mp} R_s}{R_{sh}}. \quad (6)$$

The derivative  $dI/dV$  is obtained from Equation (1) as:

$$\frac{dI}{dV} = -\left(\frac{I_{sat}}{nN_s V_{th}} \left[ \exp\left(\frac{V + IR_s}{nN_s V_{th}}\right) \right] + \frac{1}{R_{sh}}\right) \left(1 + \frac{R_s}{R_{sh}} + \frac{I_{sat} R_s}{nN_s V_{th}} \exp\left(\frac{V + IR_s}{nN_s V_{th}}\right)\right). \quad (7)$$

This expression for the derivative is valid at any point along the I-V curve of a PV module. The power derivative  $dP/dV$  at any point along the P-V curve is:

$$\frac{dP}{dV} = \frac{d}{dV}(VI) = V \frac{\partial I}{\partial V} + I \frac{\partial V}{\partial V}. \quad (8)$$

At the MPP, the derivative  $dP/dI$  is equal to zero, thus:

$$\left. \frac{dP}{dV} \right|_{\text{MPP}} = 0 = V_{mp} \frac{\partial I}{\partial V} + I_{mp}. \quad (9)$$

Consequently, at the MPP, we can write:

$$\left. \frac{dI}{dV} \right|_{\text{MPP}} = -\frac{I_{mp}}{V_{mp}}. \quad (10)$$

Therefore, at the MPP, from Equations (7) and (10), we can write:

$$\left. \frac{dP}{dV} \right|_{\text{MPP}} = -\frac{I_m}{V_m} = \frac{-\frac{1}{R_{sh}} - \frac{I_{sat}}{n N_S V_{th}} \exp\left(\frac{V_m + I_m R_s}{n N_S V_T}\right)}{1 + \frac{R_s}{R_{sh}} + \frac{I_{sat} R_s}{n N_S V_{th}} \exp\left(\frac{V_m + I_m R_s}{n N_S V_T}\right)}. \quad (11)$$

The effect of the shunt resistance is dominant at high currents and low voltages, i.e., near the short-circuit point on the I-V curve, hence we may write [17,21]:

$$\left. \frac{dI}{dV} \right|_{I_{sc}} = -\frac{1}{R_{sh}}. \quad (12)$$

The above expressions are the key equations which are used in formulating a variety of analytical [8,22], numerical, and iterative methods [3] for extracting the five parameters of the SDM model and also for formulating the parameter adjustment methods for modifying the SDM parameters according to variations in operating conditions of irradiance and temperature [19,23,24]. In addition to conventional analytical and numerical methods of parameters' extraction, there are other schemes, such as those based on the meta-heuristic firefly algorithm [25] and those based on computational intelligence, such as [26].

### 3. Method of Adjusting the SDM Model Parameters

The parameters of the SDM model of a PV module are almost always extracted at standard test conditions. Consequently, these parameters are valid only at STC. But, since the characteristics of a PV module and its model parameters vary considerably with irradiance and temperature, these extracted parameters must be adjusted to match the prevailing environmental conditions of irradiance and ambient temperature [3,9,10].

#### 3.1. Short-Circuit Current

There are two primary methods which express changes of the short-circuit current with irradiance  $G$  and temperature  $T$  as described next.

##### 3.1.1. Method 1

In this mostly reported method, the short-circuit current  $I_{sc}(G, T)$  as a function of irradiance and temperature is given as [15,22,27,28]:

$$I_{sc}(G, T) = \frac{G}{G_{STC}} [I_{sc,STC} + \mu_{I_{sc}}(T - T_{STC})] \quad (13)$$

where  $\mu_{I_{sc}}$  (A/°C) is the short-circuit current temperature coefficient, which is always available in a datasheet of a PV module. If dependence on irradiance is neglected:

$$I_{sc}(T) = [I_{sc,STC} + \mu_{I_{sc}}(T - T_{STC})]. \quad (14)$$

Similarly, if dependence on temperature is ignored:

$$I_{sc}(G) = \frac{G}{G_{STC}} I_{sc,STC}. \quad (15)$$

### 3.1.2. Method 2

This method improves the accuracy of Equation (13) by introducing an exponent to take into account the nonlinear effects of irradiance on the module's short-circuit current as [23,29,30]:

$$I_{sc}(G, T) = \left( \frac{G}{G_{STC}} \right)^\alpha [I_{sc,STC} + \mu_{I_{sc}}(T - T_{STC})] \quad (16)$$

where, the exponent  $\alpha$  is given by:

$$\alpha = \ln\left(\frac{I_{sc,STC}}{I_{sc}}\right) \Bigg/ \ln\left(\frac{G_{STC}}{G}\right). \quad (17)$$

### 3.2. Open-Circuit Voltage

Most reported studies suggest that the open-circuit voltage  $V_{OC}$  is strongly related to temperature while its dependence on irradiance has a modest effect. The dependence of the open-circuit voltage on temperature is expressed using the voltage temperature coefficient  $\mu_{V_{oc}}$  (V/°C), which is always specified in datasheets. There are five main schemes which associate the dependence of the open-circuit voltage to irradiance and temperature, as explained below.

#### 3.2.1. Method 1

This method assumes that the open-circuit voltage,  $V_{oc}(T)$  depends only on temperature  $T$  (°C) and may be estimated from [3,31]:

$$V_{oc}(T) = V_{oc,STC} + \mu_{V_{oc}}(T - T_{STC}). \quad (18)$$

#### 3.2.2. Method 2

From Equation (4), it is evident that the open-circuit voltage depends on the photocurrent, which depends on irradiance [16,17]. Hence, to include the dependence on irradiance, we can write:

$$V_{oc}(G) = \ln\left(\frac{I_{ph}(G)R_{sh} - V_{oc}(G)}{I_{sat}R_{sh}}\right) nN_s V_{th}. \quad (19)$$

These two equations, i.e., (18) and (19), can be solved numerically to obtain the dependence of the open-circuit voltage on any arbitrary temperature and irradiance.

#### 3.2.3. Method 3

Considering Equation (4) and assuming that  $R_{sh}I_{ph} \gg V_{oc}$ , the dependence of  $V_{oc}$  on the photocurrent is approximated as [2,21,32]:

$$V_{oc} = nN_s V_{th} \ln(I_{ph}/I_{sat}). \quad (20)$$

Using Equation (20), we can introduce the dependence of the open-circuit voltage on irradiance at constant temperature as:

$$V_{oc}(G) - V_{oc}(G_{STC}) = nN_s V_{th} \left[ \ln\left(\frac{I_{ph}}{I_{sat}}\right) - \ln\left(\frac{I_{ph,STC}}{I_{sat,SCT}}\right) \right]. \quad (21)$$

This may be rewritten as:

$$V_{oc}(G) - V_{oc}(G_{STC}) = nN_s V_{th} \left[ \ln\left(\frac{\frac{G}{G_{STC}} I_{ph,STC}}{I_{sat}}\right) - \ln\left(\frac{I_{ph,STC}}{I_{sat,SCT}}\right) \right]. \quad (22)$$

Assuming that the saturation current is independent of irradiance, i.e.,

$$I_{sat} = I_{sat,STC} = \text{constant (A)}, \quad (23)$$

then:

$$V_{oc}(G) - V_{oc}(G_{STC}) = \frac{N_s k T n}{q} \ln(G/G_{STC}). \quad (24)$$

The dependence of  $V_{oc}$  on both temperatures and irradiance can then be expressed as:

$$V_{oc}(G, T) = V_{oc,STC} + \frac{N_s k T n}{q} \ln(G/G_{STC}) + \mu_{V_{oc}}(T - T_{STC}). \quad (25)$$

Note that in the expression  $N_s k T n / q$ , the temperature  $T$  must be in Kelvin.

#### 3.2.4. Method 4

This method expresses the dependence of the open-circuit voltage on irradiance  $G$  and temperature  $T$  as [15,33]:

$$V_{oc}(G, T) = V_{oc,STC} + C_1 \ln(G/G_{STC}) + C_2 (\ln(G/G_{STC}))^2 + C_3 (\ln(G/G_{STC}))^3 + \mu_{V_{oc}}(T - T_{STC}) \quad (26)$$

where  $C_1 = 5.468511 \times 10^{-2}$  V,  $C_2 = 5.973869 \times 10^{-3}$  V, and  $C_3 = 7.616178 \times 10^{-4}$  V for silicon.

#### 3.2.5. Method 5

The prime advantage of this method is that it takes into account the nonlinear effects of irradiance and temperature on the open-circuit voltage,  $V_{oc}(G, T)$ , which improves accuracy [23,29,30]:

$$V_{oc}(G, T) = \frac{V_{oc,STC}}{1 + \beta \ln\left(\frac{G_{STC}}{G}\right)} \left(\frac{T_{STC}}{T}\right)^\gamma \quad (27)$$

$$\beta = [(V_{oc,STC}/V_{oc,G}) - 1] / \ln(G_{STC}/G) \quad (28)$$

$$\gamma = \ln(V_{oc,STC}/V_{oc,T}) / \ln(T/T_{STC}) \quad (29)$$

where  $V_{oc,STC}$ ,  $V_{oc,G}$ , and  $V_{oc,T}$  are the values of the open-circuit voltage at standard test conditions, at any irradiance  $G$ , and at any temperature  $T$ , respectively.

### 3.3. Photocurrent $I_{ph}$

The photocurrent is a strong function of irradiance. However, it can be expressed in terms of the lumped-circuit parameters of the single-diode model and data normally given in the module's datasheet. For example, using Equation (5), we can express the photocurrent at STC as:

$$I_{ph,STC} = \left(1 + \frac{R_s}{R_{sh}}\right) I_{sc} + I_{sat} \left[ \exp\left(\frac{I_{sc} R_s}{n N_s V_{th}}\right) - 1 \right]. \quad (30)$$

There are different schemes of expressing the photocurrent reported in the literature as examined next.

#### 3.3.1. Method 1

This method is derived from the short-circuit Equation (5), after making the assumption that under short-circuit condition, the entire photocurrent is the short-circuit current. That is, both the diode and the shunt resistance currents are too small and, hence, can be neglected [20]. This means the short-circuit equation is reduced to [2,15,32,33]:

$$I_{ph} = I_{sc}. \quad (31)$$

Equation (31), informs us that the effects of the irradiance and temperature on the photocurrent are the same as those on the short-circuit current. Consequently, Equations (14), (16), and (17) can also be used to estimate the value of the photocurrent  $I_{ph}$  under any arbitrary values of temperature and irradiance.

### 3.3.2. Method 2

This method is based on the short-circuit Equation (5) and neglects the diode current, however, the shunt resistance current is not neglected, hence [3,24]:

$$I_{ph} = \left(1 + \frac{R_s}{R_{sh}}\right) I_{sc}. \quad (32)$$

### 3.3.3. Method 3

This method is based on the open-circuit voltage Equation (3), in which the exponential term is much larger than unity. Thus, the  $(-1)$  can be neglected [20]. Consequently, the photocurrent may be written as:

$$I_{ph} = I_{sat} \exp\left(\frac{V_{oc}}{nN_s V_{th}}\right) + \frac{V_{oc}}{R_{sh}}. \quad (33)$$

The photocurrent temperature dependence can now be introduced into Equation (33) as [16,17]:

$$I_{ph}(T) = I_{sat}(T) \exp\left[\left(\frac{V_{oc,STC} + \mu V_{oc}(T - T_{STC})}{nN_s V_{th}(T)}\right)\right] + \left(\frac{V_{oc,STC} + \mu V_{oc}(T - T_{STC})}{R_{sh}}\right). \quad (34)$$

The variation of the photocurrent with irradiance and temperature is:

$$I_{ph}(G, T) = \frac{G}{G_{STC}} I_{ph}(T). \quad (35)$$

### 3.3.4. Method 4

Upon combining Equations (3) and (5), we can readily derive the following expression for the photocurrent at STC [21]:

$$I_{ph,STC} = \frac{\left[\left(1 + \frac{R_s}{R_{sh}}\right) I_{sc,STC} \left[\exp\left(\frac{V_{oc,STC}}{nN_s V_{th}}\right) - 1\right] + \frac{V_{oc,STC}}{R_{sh}} \left[1 - \exp\left(\frac{I_{sc,STC} R_s}{nN_s V_{th}}\right)\right]\right]}{\left(\exp\left(\frac{V_{oc,STC}}{nN_s V_{th}}\right) - \exp\left(\frac{I_{sc,STC} R_s}{nN_s V_{th}}\right)\right)}. \quad (36)$$

For any arbitrary conditions of temperature  $T$  and irradiance  $G$ , the photocurrent can be estimated from:

$$I_{ph}(G, T) = \frac{G}{G_{STC}} [I_{ph,STC} + \mu I_{STC}(T - T_{STC})]. \quad (37)$$

## 3.4. Ideality Factor $n$

The ideality factor,  $n$ , is an important parameter that describes how faithfully a diode characteristic follows the ideal diode equation [34]. Typically, for a silicon PV cell, the value of the ideality factor is between 1 and 2 and, normally, the ideality factor is assumed independent of both irradiance and temperature, i.e., remains constant [3,16,21].

## 3.5. Saturation Current $I_{sat}$

Most studies in the literature state that the saturation current, whose value dictates the shape of the I-V curve around the MPP, depends on temperature and irradiance [9,10,35]. Different schemes have been proposed to express its variation with changes in temperature and irradiance, and the five most-reported methods are studied next.

### 3.5.1. Method 1

In this scheme, it is assumed that the saturation current depends only on temperature and it is derived using the OC Equation (3) and the SC Equation (5). In addition, this scheme uses some information that is always available in the datasheet of a PV module, e.g., the short-circuit current and the open-circuit voltage. The variation of the saturation current with temperature,  $T$  ( $^{\circ}\text{C}$ ), is expressed as [16,21]:

$$I_{sat}(T) = \left( I_{sc}(T) - \frac{V_{oc}(T) - I_{sc}(T)R_s}{R_{sh}} \right) \exp\left(\frac{-V_{oc}(T)}{nN_s V_{th}(T)}\right). \quad (38)$$

Hence, upon substituting for  $I_{sc}(T)$  from Equation (14) and for  $V_{OC}(T)$  from Equation (18), we can write the saturation current as:

$$I_{sat}(T) = \frac{I_{sc,STC} + \mu_{I_{sc}}(T - T_{STC}) - \frac{[V_{oc,STC} + \mu_{V_{oc}}(T - T_{STC}) - (I_{sc,STC} + \mu_{I_{sc}}(T - T_{STC}))R_s]}{R_{sh}}}{\exp\left(q \frac{V_{oc,STC} + \mu_{V_{oc}}(T - T_{STC})}{n N_s k T}\right)}. \quad (39)$$

### 3.5.2. Method 2

This scheme assumes that the saturation current is independent of irradiance. In addition, the scheme assumes that the shunt resistance is high enough to be considered as an open-circuit and the photocurrent is the same as the short-circuit current. Hence, from Equation (3), we can write [3,36,37]:

$$I_{sat}(T) = \frac{I_{sc}(T)}{\exp\left(\frac{qV_{oc}(T)}{nN_s k T}\right) - 1}. \quad (40)$$

Substituting for the short-circuit current from Equation (14) and for the open-circuit voltage from Equation (18), we can write the saturation current as:

$$I_{sat}(T) = \frac{I_{sc,STC} + \mu_{I_{sc}}(T - T_{STC})}{\exp\left(q \frac{V_{oc,STC} + \mu_{V_{oc}}(T - T_{STC})}{n N_s k T}\right) - 1}. \quad (41)$$

### 3.5.3. Method 3

This scheme also assumes that the saturation current depends only on temperature and it is related to the bandgap energy  $E_g$  and temperature as [24,31,38,39]:

$$I_{sat}(T) = CT^3 \exp\left(-\frac{qE_g}{nkT}\right) \quad (42)$$

where  $C$  is a constant that depends on the diffusion properties of the junction and temperature  $T$  is in Kelvin. Therefore, the saturation current at STC is:

$$I_{sat,STC} = CT_{STC}^3 \exp\left(-\frac{qE_g}{nkT_{STC}}\right). \quad (43)$$

Dividing Equation (43) by (42), we obtain an expression for the saturation current at any arbitrary temperature,  $T$ , as:

$$I_{sat}(T) = I_{sat,STC} \left(\frac{T}{T_{STC}}\right)^3 \exp\left[\frac{qE_g}{nk} \left(\frac{1}{T_{STC}} - \frac{1}{T}\right)\right]. \quad (44)$$



## 3.5.4. Method 4

Substituting Equation (3) into (5):

$$I_{sc} = I_{sat} \left( \exp\left(\frac{V_{oc}}{nN_s V_{th}}\right) - 1 \right) - \left[ \exp\left(\frac{I_{sc} R_s}{nN_s V_{th}}\right) - 1 \right] - \frac{I_{sc} R_s}{R_{sh}} + \frac{V_{oc}}{R_{sh}}. \quad (45)$$

Neglecting the  $(-1)$  in the two exponential terms and re-arranging [21],

$$I_{sat} = \frac{\left(1 + \frac{R_s}{R_{sh}}\right) I_{sc} - \frac{V_{oc}}{R_{sh}}}{\left(\exp\left(\frac{V_{oc}}{nN_s V_{th}}\right) - \exp\left(\frac{I_{sc} R_s}{nN_s V_{th}}\right)\right)}. \quad (46)$$

By introducing the temperature dependence of  $I_{sc}$  from Equation (14) and the dependence of the open-circuit voltage on temperature and irradiance from Equation (25), we can deduce that

$$I_{sat} = \frac{(1 + R_s/R_{sh})(I_{sc,STC} + \mu_{Isc}(T - T_{STC})) - [V_{oc,STC} + \mu_{Voc}(T - T_{STC}) + \frac{nN_s kT}{q} \ln(G/G_{STC})]/R_{sh}}{\exp\left(\frac{V_{oc,STC} + \mu_{Voc}(T - T_{STC}) + \frac{nN_s kT}{q} \ln(G/G_{STC})}{n N_s kT/q}\right) - \exp\left(\frac{(I_{sc,STC} + \mu_{Isc}(T - T_{STC})) R_s}{n N_s kT/q}\right)}. \quad (47)$$

## 3.5.5. Method 5

After making the assumption that the shunt resistance is infinitely large, the saturation current can be obtained from the open-circuit voltage Equation (3) at STC as follows [32,40]:

$$I_{sat,STC} = \frac{I_{ph,STC}}{\exp(V_{oc}/nN_s V_{th}) - 1}. \quad (48)$$

Solving for the open-circuit voltage:

$$V_{oc} = \frac{nN_s kT}{q} \ln\left(\frac{I_{ph,STC}}{I_{sat,STC}} + 1\right). \quad (49)$$

We also have [2,21]:

$$V_{oc}(G, T) - V_{oc}(G, T_{STC}) = -|\mu_{Voc}|(T - T_{STC}) \quad (50)$$

and

$$I_{ph}(G, T) = \alpha_G (I_{sc,STC} + \mu_{Isc}(T - T_{STC})) \quad (51)$$

where

$$\alpha_G = G/G_{STC}. \quad (52)$$

Substituting Equation (49) into Equation (50) and substituting for the thermal voltage from Equation (2):

$$\frac{N_s k n}{q} \left[ T \ln\left(\frac{I_{ph}}{I_{sat}} + 1\right) - T_{STC} \ln\left(\frac{I_{ph,STC}}{I_{sat,STC}} + 1\right) \right] = -|\mu_{Voc}|(T - T_{STC}). \quad (53)$$

Using Equation (51), we can rewrite Equation (53) as:

$$\frac{N_s k n}{q} \left[ T \ln\left(\alpha_G \frac{(I_{sc,STC} + \mu_{Isc}(T - T_{STC}))}{I_{sat}} + 1\right) - T_{STC} \ln\left(\frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1\right) \right] = -|\mu_{Voc}|(T - T_{STC}) \quad (54)$$

$$\left[ \ln\left(\alpha_G \frac{(I_{sc,STC} + \mu_{Isc}(T - T_{STC}))}{I_{sat}} + 1\right) - \frac{T_{STC}}{T} \ln\left(\frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1\right) \right] = -\frac{q|\mu_{Voc}|(T - T_{STC})}{N_s k n T} \quad (55)$$

$$-\frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} = \ln \left[ \alpha_G \frac{(I_{sc,STC} + \mu_{Isc}(T-T_{STC}))}{I_{sat}} + 1 \right] - \ln \left[ \left( \frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1 \right)^{\frac{T_{STC}}{T}} \right] \quad (56)$$

$$\frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} = -\ln \left[ \left( \alpha_G \frac{(I_{sc,STC} + \mu_{Isc}(T-T_{STC}))}{I_{sat}} + 1 \right) \left/ \left( \frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1 \right)^{\frac{T_{STC}}{T}} \right. \right]. \quad (57)$$

Therefore:

$$\frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} = \ln \left[ \left( \frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1 \right)^{\frac{T_{STC}}{T}} \left/ \left( \frac{\alpha_G (I_{sc,STC} + \mu_{Isc}(T-T_{STC}))}{I_{sat}} + 1 \right) \right. \right] \quad (58)$$

and

$$\exp \left( \frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} \right) = \left( \frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1 \right)^{\frac{T_{STC}}{T}} \left/ \left( \alpha_G \frac{(I_{sc,STC} + \mu_{Isc}(T-T_{STC}))}{I_{sat}} + 1 \right) \right. \quad (59)$$

$$\left( \frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1 \right)^{\frac{T_{STC}}{T}} = \alpha_G \frac{(I_{sc,STC} + \mu_{Isc}(T-T_{STC}))}{I_{sat}} \exp \left( \frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} \right) + \exp \left( \frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} \right) \quad (60)$$

$$\left( \frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1 \right)^{\frac{T_{STC}}{T}} - \exp \left[ \frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} \right] = \left[ \frac{\alpha_G (I_{sc,STC} + \mu_{Isc}(T-T_{STC}))}{I_{sat}} \right] \exp \left[ \frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T} \right]. \quad (61)$$

Simplifying, the saturation current as a function of irradiance and temperature is given by [32,40]:

$$I_{sat}(G, T) = \frac{[(\alpha_G (I_{sc,STC} + \mu_{Isc}(T-T_{STC})) \exp(\frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T}))]}{[(\frac{\alpha_G I_{sc,STC}}{I_{sat,STC}} + 1)^{\frac{T_{STC}}{T}} - \exp(\frac{q|\mu_{Voc}|(T-T_{STC})}{N_s k n T})]}. \quad (62)$$

### 3.6. Series and Shunt Resistances $R_s, R_{sh}$

The series resistance is an important parameter which can have adverse effects on the power yield of a PV module and its fill factor [15,23]. The shunt resistance models the p-n junction non-idealities and its effect is equivalent to partial short-circuiting at the junction, which reduces the output current. There are four commonly reported schemes that describe the effects of irradiance and temperature on these resistances, as illustrated next.

#### 3.6.1. Method 1

Several reported studies in the literature neglect the effects of irradiance and temperature on the both resistances [3,17,21,28,32]. That is:

$$\begin{aligned} R_s(G, T) &= R_{s,STC} = \text{Constant} \\ R_{sh}(G, T) &= R_{sh,STC} = \text{Constant} \end{aligned} \quad (63)$$

#### 3.6.2. Method 2

In this method, the series resistance is independent of both temperature and irradiance, while the shunt resistance varies inversely, only, with irradiance, according to [9,31]:

$$\begin{aligned} R_s(G, T) &= R_{s,STC} \\ R_{sh}(G) &= (G_{STC}/G)R_{sh,STC} \end{aligned} \quad (64)$$

### 3.6.3. Method 3

This method is based on the assumption that both resistances vary inversely with irradiance but both are independent of temperature [15,33,41]:

$$\begin{aligned} R_s(G) &= (G_{STC}/G)R_{s,STC} \\ R_{sh}(G) &= (G_{STC}/G)R_{sh,STC} \end{aligned} \quad (65)$$

### 3.6.4. Method 4

In this scheme, it is assumed that the series resistance increases with temperature and decreases with irradiance. The shunt resistance decreases with irradiance, but is independent of temperature [38,42]:

$$\begin{aligned} R_s(G, T) &= \frac{T}{T_{STC}}(1 - \lambda \ln(G/G_{STC})) R_{s,STC} \\ R_{sh}(G, T) &= (G_{STC}/G)R_{sh,STC} \end{aligned} \quad (66)$$

where  $\lambda$  is a dimensionless constant whose value is 0.217 [38].

## 4. Results and Discussion

To evaluate and compare the performance of the parameter adjustment methods discussed in the previous section, an extensive investigation was carried out, which involved testing these methods using three PV modules of different technologies. These modules were the monocrystalline shell SQ150 [43], the multicrystalline KC175GT [44], and the thin film shell ST40 [45]. The specifications of these modules are summarized in Table 1. The constants,  $\alpha$ ,  $\beta$ , and  $\gamma$ , which were estimated using Equations (17), (28), and (29) along with information from the datasheets, are recorded in Table 2.

**Table 1.** Summary of datasheet parameters of the PV modules used in this work at STC.

Parameters	Shell SQ150	KC175GT	Shell ST40
$I_{sc}$	4.8 A	8.09 A	2.68 A
$V_{oc}$	$4.34 \times 10$ V	$2.92 \times 10$ V	$2.33 \times 10$ V
$I_{mp}$	4.4 A	7.42 A	2.41 A
$V_{mp}$	$3.4 \times 10$ V	$2.36 \times 10$ V	$1.66 \times 10$ V
$\mu_{V_{oc}}$ (V/°C)	$-161 \times 10^{-3}$	$-1.09 \times 10^{-1}$	$-100 \times 10^{-3}$
$\mu_{I_{sc}}$ (A/°C)	$1.4 \times 10^{-3}$	$3.18 \times 10^{-3}$	$0.35 \times 10^{-3}$
$N_s$	72	84	36

**Table 2.** Estimated values of the constants for the test PV modules.

Constant	$\alpha$	$\beta$	$\gamma$
Shell SQ150	0.998	0.055	1.0797
KC175GT	0.977	0.053	1.32
Shell ST40	0.996	0.085	1.367

The reference values used to compare the parameter adjustment methods were extracted using the well-verified numerical method based on Newton–Raphson algorithm [28]. These are shown in Table 3 for the three test modules.

Table 4 shows the measured and calculated values, using Methods 1 and 2, of the variations of the short-circuit current with irradiance at a temperature of 25 °C. The percentage relative errors between the calculated and measured values, which are also included in the table, indicate that both methods yielded relatively close results. However, Method 2 was more accurate at higher irradiance level since it took into account the nonlinear dependence of the short-circuit current on irradiance.

**Table 3.** Estimated values of the parameters of the test PV module.

Parameters	Shell SQ150	KC175GT	Shell ST40
$n$	1.4397	1.5036	1.5028
$R_s$	$5.906 \times 10^{-1} \Omega$	$1.061 \times 10^{-1} \Omega$	1.4226 $\Omega$
$R_{sh}$	$1.1661 \times 10^3 \Omega$	$3.251018 \times 10^2 \Omega$	952.405 $\Omega$
$I_{sat}$	$4.0163 \times 10^{-7} \text{ A}$	$1.1662 \times 10^{-6} \text{ A}$	$1.4057 \times 10^{-7} \text{ A}$
$I_{ph}$	4.8024 A	8.0926 A	2.684 A

**Table 4.** Measured and calculated values (by Methods 1 and 2, Section 3.1) of the short-circuit current  $I_{sc}$  (A) and the relative error.

PV Module	$I_{sc}$ (A)	Irradiance W/m <sup>2</sup>				
		1000	800	600	400	200
KC175GT	Measured	8.09	6.80889	4.91094	3.27396	1.56581
	Method 1	8.09	6.472	4.854	3.236	1.618
	%error		4.95478	1.1595	1.1595	3.331
	Method 2	8.09	6.5053	4.9114	3.3049	1.679
	%error		4.4587	0.0094	0.945	7.23
SQ150	Measured	4.8	3.84	2.88	1.90884	0.94884
	Method 1	4.8	3.84	2.88	1.92	0.96
	%error		0	0	0.5847	1.1762
	Method 2	4.8	3.8417	2.8829	1.9235	0.9631
	%error		0.0443	0.1007	0.768	1.5029
ST40	Measured	2.68	2.14894	1.61171	1.07447	0.53724
	Method 1	2.68	2.144	1.608	1.072	0.536
	%error		0.2299	0.2302	0.2299	0.2301
	Method 2	2.68	2.1459	1.6113	1.0759	0.539
	%error		0.1415	0.0254	0.1331	0.4207

Table 5 shows the values of the photocurrent calculated using Equation (30) and calculated using the parameter adjustment methods in Section 3.3. It was evident from the relative errors, that the photocurrent calculations were similar for all methods. This confirmed that the photocurrent dependence on temperature was very modest. However, it is worth noting that Method 4 (for the photocurrent Section 3.3.4) matched the results of Equation (30) as it does not involve the approximations, which the other method used.

**Table 5.** Calculated photocurrent (A) using different methods.

PV Modules	Equation (30)	Method 1	Method 2	Method 3	Method 4
KC175GT	8.0926406	8.09	8.09263991	8.09264031	8.0926406
SQ150	4.8024316	4.8	4.80243089	4.80243114	4.80243165
ST40	2.6840051	2.68	2.6840031	2.68400152	2.68400513

Tables 6–8 compare the measured values of the open-circuit voltage with those calculated at various levels of irradiance using Methods 1 to 5 discussed in Section 3.2 (for the open-circuit voltage Sections 3.2.1–3.2.5) for the three test modules. It was evident that Method 5 outperformed all other methods, which confirmed that the relationship between the open-circuit voltage and irradiance involved nonlinear dependence. Method 4 required a high degree of accuracy in measuring data under different levels of irradiance. Methods 2 and 3 had similar results due to the similarities of the assumptions involved in deriving both methods.

**Table 6.** Calculated and measured open-circuit voltage  $V_{oc}$  (V) and relative error for the shell SQ150 module under different irradiance levels  $G$  ( $W/m^2$ ) and temperature of 25 °C.

$G$ ( $W/m^2$ )	Measured	Method 1	Method 2	Method 3	Method 4	Method 5
1000	43.4	43.4	43.4	43.4	43.4	43.4
%error		0	0	0	0	0
800	42.91547	43.4	42.823523	42.80548	43.38809	42.87381
%error		1.1165	0.2143	0.2563	1.1013	0.0971
600	42.22329	43.4	42.05706	42.03902	43.37352	42.21398
%error		2.7869	0.3937	0.4364	2.7242	0.0221
400	41.25423	43.4	40.97679	40.95875	43.35267	41.31775
%error		5.2013	0.6725	0.7164	5.0866	0.154
200	39.59298	43.4	39.13005	39.11201	43.32139	39.87068
%error		8.7719	1.1692	1.2148	9.4169	0.70139

**Table 7.** Calculated and measured  $V_{oc}$  (V) and relative error of the KC175GT module under different irradiance levels and a temperature of 25 °C.

Irradiance		$V_{oc}$ (V)				
( $W/m^2$ )	Measured	Method 1	Method 2	Method 3	Method 4	Method 5
1000	29.2	29.2	29.2	29.2	29.2	29.2
%error		0	0	0	0	0
800	28.81579	29.2	28.80708	28.78606	29.18809	28.8587
%error		1.365	0.03023	0.1032	1.292	0.1489
600	28.43158	29.2	28.27343	28.25241	29.17352	28.43029
%error		2.7027	0.5563	0.6302	2.6096	0.4554
400	27.81684	29.2	27.52128	27.50026	29.15432	27.8476
%error		4.9724	1.0625	1.1381	4.8082	0.1107
200	27.04842	29.2	26.23548	26.21446	29.12429	26.905
%error		7.9546	3.0055	3.0832	7.6763	0.5302

**Table 8.** The calculated and measured  $V_{oc}$  (V) and relative error of the shell ST40 module under different irradiance levels and a temperature of 25 °C.

Irradiance		$V_{oc}$ (V)				
( $W/m^2$ )	Measured	Method 1	Method 2	Method 3	Method 4	Method 5
1000	23.3	23.3	23.3	23.3	23.3	23.3
%error		0	0	0	0	0
800	22.79815	23.3	22.99962	22.98971	23.28809	22.86629
%error		2.2013	0.8837	0.8402	2.149	0.2989
600	22.29631	23.3	22.59959	22.58968	23.27352	22.33041
%error		4.5016	1.3602	1.3158	4.3828	0.1529
400	21.54354	23.3	22.03578	22.02587	23.25432	21.61641
%error		8.1531	2.2849	2.2389	7.941	0.3383
200	20.21723	23.3	21.07194	21.06204	23.22429	20.49609
%error		15.248	4.2276	4.1787	14.8738	1.3793

Tables 9–11 compare the measured values of the open-circuit voltage with those calculated at various levels of temperature using the methods discussed in Sections 3.2.1–3.2.5. Method 5 (Section 3.2.5) outperformed all other methods for the monocrystalline SQ150 module, while, for the other two modules, both Methods 1 and 5 had similar results. Obtaining an accurate value of open-circuit voltage with respect to temperature changes hinges upon obtaining accurate values of the temperature coefficient and the constant  $\gamma$  [23].

**Table 9.** The calculated (Methods 1 and 5) and measured  $V_{oc}$  (V) and relative error of the shell SQ150 module under different temperatures and an irradiance of 1000 W/m<sup>2</sup>.

Temperature		$V_{oc}$ (V)			
(°C)	Measured	Method 1	%Error	Method 5	%Error
20	44.205	44.205	0	44.2002	0.0109
25	43.4	43.4	0	43.4	0
30	42.7315	43.3195	1.376	42.6273	0.2439
40	41.258	40.985	0.6617	41.1587	0.2407
50	39.7845	39.375	1.0218	39.7846	0.00025
60	38.311	37.765	1.4252	38.4962	0.4834

**Table 10.** Calculated and measured  $V_{oc}$  (V), and relative error of the KC175GT module under different temperatures and an irradiance of 1000 W/m<sup>2</sup>.

Temperature		$V_{oc}$ (V)			
(°C)	Measured	Method 1	%Error	Method 5	%Error
25	29.2	29.2	0	29.2	0
50	26.26533	26.475	0.7983	26.2649	0.00164
75	23.25729	23.75	2.1185	23.8122	2.386

**Table 11.** Calculated and measured  $V_{oc}$  (V), and relative error of the shell ST40 module under different temperatures and an irradiance of 1000 W/m<sup>2</sup>.

Temperature		$V_{oc}$ (V)			
(°C)	Measured	Method 1	%Error	Method 5	%Error
20	23.8	23.8	0	23.8452	0.1899
25	23.3	23.3	0	23.3	0
30	22.81138	22.8	0.04989	22.776	0.1551
40	21.85938	21.8	0.2716	21.7872	0.3302
50	20.87077	20.8	0.3391	20.8703	0.0023
60	19.91877	19.8	0.5963	20.0184	0.5002

## 5. Conclusions

This paper presented an investigation into various methods reported in the literature for modifying the STC parameters of a PV module to suit the current operating conditions of irradiance and temperature. The STC parameters were extracted using numerical methods, which have been proven to be most accurate, and then the modification methods were applied using MATLAB-based simulation software to investigate their variations with temperature and irradiance.

Several schemes for modifying each parameter of the single-diode model with variations in temperature and irradiance were explored in Section 3. For each parameter, these schemes were referred to as Method 1, Method 2, etc., as listed in Section 3. The ideality factor was assumed constant.

The results of applying these methods were discussed in Section 4 and compared with measured values in Tables 4–11. When studying the effect of irradiance, on the short-circuit current, Method 2 (Section 3.2) was most accurate, which confirmed the nonlinear dependence of the short-circuit current on irradiance. This was particularly noticeable at higher levels of irradiance. The difference in the relative errors for all test modules was similar. The study also indicated that temperature had very modest effect on the short-circuit current.

For the open-circuit voltage, the study confirmed the nonlinear dependence of the open-circuit voltage on irradiance and showed that Methods 2 and 3 had similar relative error for all test modules, which can be explained by the fact that they were derived using similar assumptions.

The results of Methods 1 and 4 showed that ignoring the dependence of the open-circuit voltage on irradiance can lead to higher relative errors, particularly at low levels of irradiance where the nonlinear effects are more prominent. Method 2 resulted in smaller relative error when tested with the monocrystalline panel compared to the multicrystalline and thin film panels. However, the differences were modest. It is suggested that further extensive investigation should explore the dependence and suitability of the parameter adjustment methods on the technology and different materials of PV modules.

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