



Investigating a Detection Method for Viruses and Pathogens Using a Dual-Microcantilever Sensor

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Abstract: Piezoresistive microcantilever sensors for the detection of viruses, pathogens, and trace chemical gasses, with appropriate measurement and signal processing methods, can be a powerful instrument with high speed and sensitivity, with in situ and real-time capabilities. This paper discusses a novel method for mass sensing on the order of a few femtograms, using a dual-microcantilever piezoresistive sensor with a vibrating common base. The two microcantilevers have controllably shifted natural frequencies with only one of them being active. Two active piezoresistors are located on the surfaces of each of the two flexures, which are specifically connected in a Wheatstone bridge with two more equivalent passive resistors located on the sensor base. A dedicated experimental system measures the voltages of the two half-bridges and, after determining their amplitude-frequency responses, finds the modulus of their differences. The modified amplitude-frequency response possesses a cusp point which is a function of the natural frequencies of the microcantilevers. The signal processing theory is derived, and experiments are carried out on the temperature variation in the natural frequency of the active microcantilever. Theoretical and experimental data of the temperature-frequency influence and equivalent mass with the same impact are obtained. The results confirm the sensor's applicability for the detection of ultra-small objects, including early diagnosis and prediction in microbiology, for example, for the presence of SARS-CoV-2 virus, other viruses, and pathogens. The versatile nature of the method makes it applicable to other fields such as medicine, chemistry, and ecology.

Keywords: microcantilever; piezoresistor; vibration; virus detection; SARS-CoV-2

1. Introduction

The detection of chemical substances, viruses, and pathogens with masses in the femtogram range when performed by microcantilever sensors is distinguished by their low power consumption, high sensitivity, and fast response. An undoubted advantage of these sensors is that, under appropriate conditions, the detection of the chemical or the diagnosis of the presence of a particular virus can be made at an early stage, in situ in real time, even before the contamination process or harmful chemical saturation has occurred.

Single microcantilever sensors have already been applied in various fields, for example, to detect C-reactive protein (CRP) [1] or to determine optical bias for single-nucleotide mismatch recognition [2]. The simplified design of these sensors is a prerequisite for their lower cost and low power consumption and creates opportunities for the application of simple electronic circuits and signal processing software. On the other hand, compared to dual-microcantilever sensors, they have lower temperature compensation and lower sensitivity and reliability.



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Copyright: © 2024 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https:// creativecommons.org/licenses/by/ 4.0/). In dual-microcantilever sensors, one of the most popular detection methods is based on comparing the natural frequencies of a passive and active microcantilever. Tian X. et al. demonstrated dual-microcantilever sensors containing a microcantilever for hydrogen sulfide gas detection, which compared with metal oxide gas sensors have ultra-low power and high sensitivity in certain cases [3]. The ratio of amplitudes in a coupled dualmicrocantilever beam has been used as a sensing factor by achieving noise suppression, smaller damping forces, and the larger mass difference of the microcantilevers [4]. A dual-microcantilever sensor with high sensitivity and trace hydrogen sulfide gas detection capabilities with positive and negative frequency shifts has been demonstrated in [5]. In some studies on the application of piezoelectric polymer two-microcantilever sensors, it has been shown that higher sensitivity is achieved when operating with the second order of natural frequencies [6].

The operating principles of microcantilever piezoresistive sensors are mainly divided into static and dynamic. In the former, the signal is produced by the static deformation of the microcantilever, which affects the parameters of an electrical circuit through the piezoresistive effect [7,8]. In dynamic methods, the microcantilever is forced to vibrate, for example, by photo- or electro-thermal effects [9,10], by the vibratory actuation of the substrate [11,12], or by other actuation characteristics of microelectromechanical systems (MEMS), such as piezoelectric [13], capacitive [14], or magnetic actuation [15].

In terms of the application of microcantilever sensors, more versatile applications are envisaged beyond those in microbiology or chemistry. Genesensors obtained by modifying the surface of a microcantilever with applications in biology, chemistry, pharmaceutics, and environmental monitoring have been systematically studied in terms of immobilization processes, complementary hybridization, and signal extraction and processing [16].

The theory for the study of microcantilever sensors includes both lumped and distributed dynamic models. It is well known that distributed parameter models give a clearer picture of the behavior of the objects under study, but due to their complexity, they pose some theoretical difficulties, for example, in describing resonant modes. A dynamic distributed parameter model of a cantilever with base excitation and tip mass was presented by To C. in [17], and studies of forced vibrations near the resonant mode were modeled by Repetto et al. in [18]. Alternatively, lumped parameter models find applications in solving a variety of problems, such as those related to antistiction problems [19] or for static and modal analysis [20].

The aim of this paper is to elucidate and experimentally validate the theoretical basis of a new method for the detection of viruses, pathogens, and chemical gasses with masses in the order of several femtograms, based on the measurement of the frequency of a cusp point in the amplitude–frequency response of two microcantilevers. The research presented here summarizes some of the work of several research projects and is based on a recent patent-applied method [21]. In a publication by Banchelli et al. [22], the patented method was investigated with respect to its robustness and sustainability.

2. Sensor Description and Problem Statement Formulation

The sensor consists of two silicon microcantilevers with a common base. One piezoresistor with a resistance R_0 is located on the surface of each of the microcantilevers and two equivalent resistors are added at the base. Thin-film aluminum U-shaped heaters were fabricated on the two microcantilevers, as shown in Figure 1a. The piezoresistors and passive resistors were connected in a Wheatstone bridge, as shown in Figure 1b. Through the two passive resistors, temperature compensation is achieved in the output signal of the Wheatstone bridge. Figure 1c reveals the appearance of the sensor. The close-up of the sensor shown in Figure 1d shows that thin-film gold-coated pads are patterned on the two microcantilevers, and each can be selectively activated for detection, while the other performs a passive function. The gold pads are not shown in Figure 1a.



Figure 1. Schematic overview and topology of dual-cantilever microsensor: (**a**) electromechanical schematic; (**b**) Wheatstone bridge circuit; (**c**) photo of sensor; (**d**) close-up view of sensor topology.

The principle of operation of the sensor is similar in operation to most dual-microcantilever sensors with an active and passive microcantilever but differs in the method of signal detection. The two microcantilevers have close but controllably shifted natural frequencies. The base of the two cantilevers vibrates with a monotonically varying frequency swept in a narrow range around the natural frequencies of the sensor. During the sweep at the frequency from the Wheatstone bridge, a signal is measured that has a point in its amplitude–frequency response curve between the two natural frequencies of the microcantilevers with an amplitude lower than the sensor noise level. The change in frequency of the zero point is sensitive enough to register a change in the mass of one of the cantilevers relative to the other on the order of femtograms. This principle will be elucidated in detail later.

3. Load and Dynamic Model of Cantilever Beam with Harmonic Base Excitation

The base of the microcantilevers is assumed to move by a harmonic function y_1 of the form

$$y_1 = a \sin \omega t, \tag{1}$$

where *a* is the amplitude of the excitation function, ω is the circular frequency, and *t* is the time. The axis *x* of the absolute coordinate system *Axy* is assumed to coincide with the neutral longitudinal line of the beam, and the deformation transverse displacements are performed along the axis *y* (Figure 2a). The dynamic model with lumped microcantilever parameters is depicted in Figure 2b.

The microcantilever has a length l_1 and a constant rectangular cross section with width l_2 and height l_3 .

Consider an elementary cantilever volume of length dx, located at an arbitrary distance x from the fixation point A. On this volume due to the acceleration of the base,

$$\frac{d^2 y_1}{dt} = -a\omega^2 \sin \omega t \tag{2}$$

an inertial elementary force

$$d\phi = -\left(-a\omega^2\sin\omega t\right)dm\tag{3}$$

acts with the help of the elemental mass dm, which is calculated by

$$dm = \rho l_2 l_3 dx,\tag{4}$$

where ρ is the density of silicon, the material from which the microcantilever is made.



Figure 2. Microcantilever beam with harmonically driven base: (**a**) microcantilever diagram; (**b**) lumped dynamic model of microcantilever beam.

The inertial elementary force is distributed along the length of the microcantilever with the longitudinally distributed load

$$q_i = \frac{d\phi}{dx} = \rho a l_2 l_3 \omega^2 \sin \omega t.$$
(5)

In addition, the regularly distributed weight q_G acts on the cantilever:

$$q_G = \frac{dG}{dx} = -\rho a l_2 l_3 g,\tag{6}$$

where

$$dG = -g\rho l_2 l_3 dx \tag{7}$$

is the weight of the elementary volume, and *g* is the gravity acceleration.

The two longitudinally uniformly distributed loads have equivalent concentrated inertial Q_i and weight *G* forces, which are located in the middle of the cantilever and, respectively, have the form

$$Q_i = q_i l_1 = \rho a l_1 l_2 l_3 \omega^2 \sin \omega t = m a \omega^2 \sin \omega t, \tag{8}$$

$$G = q_G l_1 = -g\rho l_1 l_2 l_3 dx. \tag{9}$$

Since the cantilever vibrates at frequencies as high as 50 kHz, and the amplitude of the vibrations is greater than 1 μ m, the acceleration generated by the base motion exceeds the ground acceleration by at least two orders of magnitude, warranting the neglection of the gravitational force. Another reason for neglecting this loading is that it causes deformations of the microcantilever on the order of several angstroms (Å) [23].

The microcantilever loaded as such, also referred to as the original, is fitted with a concentrated model [23], which is a mass point located in the middle of the microcantilever on which the applied inertial force acts (Figure 2b). The mass point moves according to the same law as that of the midpoint on the neutral line of the cantilever. The concentrated

model has an effective mass m_e and an effective stiffness k_e , obtained under the condition of equality of the model and original energies.

The motion of the concentrated mass is described by the differential equation

$$m_e \frac{d^2 y}{dt^2} + b \frac{dy}{dt} + k_e y = Q_i \tag{10}$$

where b is the viscous resistivity coefficient.

After taking (8) into account and dividing by m_{e} , (10) is rewritten as

$$\frac{d^2y}{dt^2} + 2\beta \frac{dy}{dt} + \omega^2 y = a_e \omega^2 \sin(\omega t), \tag{11}$$

where

$$\beta = \frac{b}{2m} \tag{12}$$

is the damping coefficient,

$$\varpi = \sqrt{\frac{k}{m}} \tag{13}$$

is the natural frequency of the microcantilever, and

$$a_e = \frac{ma}{m_e} \tag{14}$$

is the effective amplitude of the forced vibration.

The solution of the Linear Differential Equation (10) is obtained as the sum of the solution of the homogeneous equation and a partial integral of the inhomogeneous equation. The solution of the homogeneous equation is damped and vanishes after a short time. This is a reason to take the solution of (11) as only the forced oscillations of the microcantilever, which, for the case under consideration, have the form

$$y = B\sin(\omega t + \psi), \tag{15}$$

where the amplitude *B* of the forced oscillations is

$$B = \frac{a\omega^2}{\sqrt{(\omega^2 - \omega^2)^2 - 4\beta^2\omega^2}}$$
(16)

and the forced vibration phase has the form

$$\psi = \frac{2\omega\beta}{\omega^2 - \omega^2}.\tag{17}$$

The piezoresistors are formed in the fixed end of the cantilever, where the mechanical stress σ is calculated using the formula

$$\sigma = \frac{M_y}{W},\tag{18}$$

where the bending moment M_{y} is determined by the expression

$$M_y = \frac{Q_i l_1}{2} = -\frac{l_1^2 q}{2} \tag{19}$$

The force Q_i is substituted according to (8), and the resisting moment W_y is calculated using the formula

$$W_y = \frac{l_2 l_3^2}{6}.$$
 (20)

The transverse displacement of the static elastic line of the microcantilever under a uniformly linear distributed load is given by

$$w(x) = \frac{qx^2}{24EI} \Big(6l_1^2 - 4l_1 x + x \Big), \tag{21}$$

where

$$I_y = \frac{l_2 l_3^3}{12} \tag{22}$$

is the moment of inertia of the cross section with respect to the axis *x*.

At the midpoint of the beam at $x = \frac{l_1}{2}$ and with the help of (21), the deflection w_2 is found

$$w_2 = w\left(\frac{l_1}{2}\right) = \frac{17l_1^*q}{384EI}$$
(23)

From the above formula, *q* is expressed and substituted into (19), and after taking into account that $w_2 \equiv y$, the relationship between the midpoint displacement and the fixed-end mechanical stress is obtained:

$$\sigma = \frac{96El_3}{17l_1^2} y(t) = \frac{96El_3}{17l_1^2} B\sin(\omega t + \psi).$$
(24)

4. Basic Concepts of the Considered Piezoresistor Detection Method

At the ends of piezoresistor 1 of the Wheatstone bridge (Figure 1b), the electrical voltage is obtained

$$u_1 = \frac{R_0 + \Delta R_1}{2R_0 + \Delta R_1} v_{cc}.$$
 (25)

Similarly, the end-to-end voltage of piezoresistor 2 is measured:

$$u_2 = \frac{R_0 + \Delta R_2}{2R_0 + \Delta R_2} v_{cc}.$$
 (26)

In Formulas (25) and (26), since silicon is an anisotropic material according to [24] and [25], the relative resistance change ΔR_1 in the piezoresistor is proportional to the mechanical stress, and for plane (100) and direction [110] the relations for the two piezoresistors, respectively, are

$$\frac{\Delta K_1}{R_0} = \pi_l (1 + \nu \pi_t) \sigma_1 = \pi_R \sigma_1$$

$$\frac{\Delta R_2}{R_0} = \pi_l (1 + \nu \pi_t) \sigma_2 = \pi_R \sigma_2$$
(27)

where in plane (100) for direction [110], $\pi_l = \frac{1}{2}(\pi_{11} + \pi_{12} + \pi_{44})$ and $\pi_t = \frac{1}{3}(\pi_{11} + 2\pi_{12} - \pi_{44})$, ν is Poisson's ratio, $\pi_R = \pi_l(1 + \nu \pi_t)$, and v_{cc} is the supply voltage.

Having considered the formulas in (27), σ_1 and σ_2 , assuming a form according to (24) for the electrical voltages of the two piezoresistors, are obtained:

$$u_{1} = \frac{1 + \delta_{R}B_{1}\sin(\omega t + \psi_{1})}{2 + \delta_{R}B_{1}\sin(\omega t + \psi_{1})}v_{cc},$$

$$u_{2} = \frac{1 + \delta_{R}B_{2}\sin(\omega t + \psi_{2})}{2 + \delta_{R}B_{2}\sin(\omega t + \psi_{2})}v_{cc},$$
(28)

where

$$\delta_R = \frac{96El_3\pi_R}{17l_1^2}$$
(29)

is called the generalized piezoresistivity coefficient.

The functions in (28) are periodic and at the time points

$$t = \frac{4\pi n + \pi - 2\psi_i}{2\omega} \quad n = 1, 2, 3... \quad i = 1, 2,$$

$$t = \frac{4\pi n - \pi - 2\psi_i}{2\omega} \quad n = 1, 2, 3... \quad i = 1, 2,$$
(30)

these have the following maximum and minimum values, respectively:

$$u_{\max i} = \frac{1 + \delta_R B_i}{2 + \delta_R B_i} v_{cc},$$

$$u_{\min i} = \frac{1 - \delta_R B_i}{2 - \delta_R B_i} v_{cc} \quad i = 1, 2.$$
(31)

It is assumed that the amplitude is the measured peak-to-peak voltage; therefore, the amplitude–frequency response function is represented by

$$V_{Ai}(\omega) = u_{\max i} - u_{\min i} = \frac{2\delta_R B_i(\omega)}{4 - \delta_R^2 B_i^2(\omega)} v_{cc} \quad i = 1, 2.$$
(32)

Figure 3 shows the plots of the amplitude–frequency characteristics of microcantilever 1 and microcantilever 2 obtained by Equation (32) and the data in Table 1.



Figure 3. Calculated amplitude–frequency responses of microcantilever 1 V_{A1} [V] and microcantilever 2 V_{A2} [V] as function of frequency *f* [Hz].

Table 1. Microcantilever geometry and materials data.

Parameter	Symbol	Value	Unit
Length of microcantilever 1	l ₁₁	$294 imes 10^{-6}$	m
Length of microcantilever 2	l ₁₂	$292 imes 10^{-6}$	m
Width of microcantilever 1	l ₂₁	$150 imes 10^{-6}$	m
Width of microcantilever 2	l ₂₂	$172 imes 10^{-6}$	m
Height of microcantilever 1	l ₃₁	$4 imes 10^{-6}$	m
Height of microcantilever 2	l ₃₂	$4 imes 10^{-6}$	m
Basis resistance of a piezoresistor	R ₀	1000	Ω
Density of the silicon	ρ	2329 *	kg/m ³
Young's modulus of the n-silicon in [110] direction	E ₁₁₀	170 *	GPa
Piezoresistivity coefficient for direction 11 n-Si	π_{11}	$-102 imes 10^{-11}$ **	Pa ^{−1}
Piezoresistivity coefficient for direction 12 n-Si	π_{12}	53×10^{-11} **	Pa ^{−1}
Piezoresistivity coefficient for direction 44 n-Si	π_{44}	$-14 imes 10^{-11}$ **	Pa^{-1}

Symbol	Value	Unit
<i>C</i> ₁₁	$165.65 \times 10^9 ***$	Pa
C ₁₂	63.94×10^9 ***	Pa
C ₄₄	79.51×10^9 ***	Pa
ω_1	10,402.535	s^{-1}
ω_2	10,568.028	s^{-1}
f_{s1}	65,361.057	Hz
f_{s2}	66,400.888	Hz
v_{cc}	8	V
a_e	$9.92 imes10^{-8}$	m
β_1	1554.755	s^{-1}
β2	1675.886	s^{-1}
	$\begin{tabular}{c c c c c c } \hline Symbol \\ \hline C_{11} \\ \hline C_{12} \\ \hline C_{44} \\ \hline & & & & \\ \hline & & & \\ \hline & & & & \\ \hline & & & &$	$\begin{tabular}{ c c c c } \hline Symbol & Value \\ \hline C_{11} & 165.65 \times 10^9 *** \\ \hline C_{12} & 63.94 \times 10^9 *** \\ \hline C_{44} & 79.51 \times 10^9 *** \\ \hline $\vee 0_1$ & 10,402.535 \\ \hline $\vee 0_2$ & 10,568.028 \\ \hline f_{s1} & 65,361.057 \\ \hline f_{s2} & 66,400.888 \\ \hline v_{cc} & 8 \\ \hline a_e & 9.92 \times 10^{-8} \\ \hline β_1 & 1554.755 \\ \hline β_2 & 1675.886 \\ \hline \end{tabular}$

Table 1. Cont.

* According to data from [26]. ** According to data from [27]. *** According to data from [28].

The output voltage of the Wheatstone bridge [29] is calculated using the formula

$$V_{out} = u_1 - u_2 = \frac{[B_1 \sin(\omega t + \psi_1) - B_2 \sin(\omega t + \psi_2)]\delta_r v_{cc}}{[2 + \delta_r B_1 \sin(\omega t + \psi_1)][2 + \delta_r B_2 \sin(\omega t + \psi_2)]'},$$
(33)

which was obtained after taking into account Formula (28) and making simplifications. For the experimental study of the output voltage amplitude, the difference is measured here:

$$V_{Aout} = V_{A1} - V_{A2} = \frac{2\delta_R(B_1 - B_2)(B_1B_2\delta_R^2 + 4)}{(\delta_R^2 B_i^2 - 4)(\delta_R^2 B_2^2 - 4)}v_{cc}.$$
(34)

The graph of the difference V_{Aout} according to (34) is plotted in Figure 4a. It is noticeable that the extreme points in the figure do not coincide with the natural frequencies of the microcantilevers.



Figure 4. The amplitude–frequency responses obtained by the voltage differences from the halfbridges of the two microcantilevers: (**a**) the difference of the amplitudes of the two microcantilevers; (**b**) the absolute value of the difference of the amplitudes of the two microcantilevers.

Devices that measure the amplitude–frequency response, such as those in [30], typically convert the voltage by an absolute value, resulting in the graph shown in Figure 4b. In this graph, the cusp point f_{cusp} , which in Figure 4a is the root of the amplitude–frequency response, is clearly visible.

The cusp point f_{cusp} has a signal that is below the noise level. Further theoretical and experimental studies will show that this point is sensitive enough to the variation in the natural frequency of one of the microcantilevers to register a mass change in the order of a few femtograms (10^{-15} g).

5. Experimental Study of the Dual-Microcantilever Sensor

In parallel with the development of this theory, an experimental test system was built to verify it. Figure 5a shows the general view of the system. The sensor signals were collected in a National Instruments PXI system with up to 2 MS/s sampling capability. Signal processing was performed using a LabVIEW 11 program capable of measuring highfrequency vibrations with frequencies up to 300 kHz and a resolution of 0.01 Hz [22]. Sensor 1 is elucidated in detail in Figure 5b, where the chip, piezoelectric actuator, and housing can be seen. High-frequency electrical sine signals were produced by a Digilent sine signal generator and fed into a piezoelectric actuator, which excited mechanical vibrations at the base of the two microcantilevers.



Figure 5. Experimental system for testing piezoresistive sensors with dual-microcantilever beams:
(a) general view; (b) closer look at sensor and its actuation. 1. Sensor. 2. NI PXI system. 3. Ammeter,
4. Digilent wave generator. 5. Potentiometers for adjusting current in microcantilever heaters.
6. Batteries to power heaters. 7. Monitor. 8. Microchip. 9. Piezoelectric actuator. 10. Sensor housing.

Initially, experiments were conducted to verify the theory derived above. For this purpose, vibrations were generated through the sine wave generator, which varied in a range from 64 kHz to 68 kHz, which includes the natural frequencies of the beams. The frequency range was divided into 400 steps, and for each step, vibrations of an order of several periods were generated. For each step, the maximum amplitude between the maximum and minimum voltages was measured for the piezoresistors of the two microcantilevers separately and the output of the Wheatstone bridge. Using the LabVIEW 11 program, the results were saved in an Excel file and then processed in Maple.

Figure 6a,b show the plots of the amplitude–frequency responses for microcantilever 1 and microcantilever 2, respectively, obtained experimentally and compared with the



theoretical results obtained above. In Figure 6a, the second peak in the experimental plot is not typical and is due to random error.

Figure 6. Theoretical and experimental plots of amplitude–frequency response: (**a**) theoretical and experimental amplitude–frequency response of microcantilever 1; (**b**) theoretical and experimental amplitude–frequency response of microcantilever 2.

Figure 7a shows the voltage differences of the two amplitude–frequency responses obtained by experiment and those from theory, plotted with a solid line. Figure 7b plots the absolute value of the Wheatstone bridge output voltage.



Figure 7. A graphical representation of the amplitude–frequency response results. The experimental results are plotted with a solid line, and theoretical results are represented by a circle symbol: (**a**) the theoretical and experimental amplitude–frequency response of the output voltage of the Wheatstone bridge; (**b**) the theoretical and experimental amplitude–frequency response of the absolute value of the output voltage of the Wheatstone bridge.

It is evident from the presented graphs that there is a consistent pattern and a reasonably good accuracy of agreement between the theoretical and experimental results. It is noteworthy that in spite of larger deviations in the extreme values, at the cusp point of Figure 7b or the root of Figure 7a, the matching of the two graphs is of higher accuracy. This indicates that the method possesses high robustness, which will be the subject of other studies.

6. An Investigation of the Sensitivity of the Method through the Possibilities of Determining the Frequency of the Cusp Point

In Equation (34), the angular frequencies, ω , ω_1 , and ω_2 [rad/s], are converted into the rotational frequencies f, f_{s1} , and f_{s2} [Hz], respectively, according to the relation

$$\omega = 2\pi f. \tag{35}$$

For the damping factors β_1 and β_2 , the substitution

$$\beta_i = 2\pi\eta_i \ i = 1, \ 2 \tag{36}$$

was used. After ω , ω_1 , ω_2 , β_1 , and β_2 are transformed according to (35) and (36), they are substituted into (34), the numerator of the difference is set to zero, and the equation for the frequency of the cusp point is obtained:

$$\pi^2 a_e^2 \delta_r^2 \left(\frac{\widetilde{B}_2}{\sqrt{\widetilde{B}_2}} - \frac{\widetilde{B}_1}{\sqrt{\widetilde{B}_1}} \right) f_{cusp}^6 + 4 \left(\frac{\widetilde{B}_1 \widetilde{B}_2}{\sqrt{\widetilde{B}_1}} - \frac{\widetilde{B}_1 \widetilde{B}_2}{\sqrt{\widetilde{B}_2}} \right) f_{cusp}^2 = 0, \tag{37}$$

where

$$\widetilde{B}_{1} = \pi^{2} \left[\left(f_{cusp}^{2} - f_{s1}^{2} \right)^{2} + 4\eta_{1}^{2} f_{cusp}^{2} \right],
\widetilde{B}_{2} = \pi^{2} \left[\left(f_{cusp}^{2} - f_{s2}^{2} \right)^{2} + 4\eta_{2}^{2} f_{cusp}^{2} \right].$$
(38)

The solution of (37) yields an important result for the studies here, by which the cusp point frequency is analytically determined:

$$f_{cusp} = \frac{\sqrt{2}}{2} \sqrt{\frac{f_{s1}^4 - f_{s2}^4}{f_{s1}^2 - f_{s2}^2 - 2(\eta_1^2 - \eta_2^2)}}.$$
(39)

By the formula thus derived, it is seen that the dependence of the cusp point frequency on the natural frequencies and damping of the microcantilevers can be analytically investigated.

For the case under consideration and from the data in Table 1, the cusp point in Equation (39) was calculated to be $f_{cusp} = 65,889.06309$ Hz, which confirms the graphical results for the cusp point in the theory and experiments obtained in Figures 4b and 7b. The same result was confirmed by numerically solving the equation directly composed from (34).

Formula (39) shows that f_{cusp} depends only on the natural frequencies and damping of the system, indicating that the method is stable and is not affected by the amplitude of the excitation base vibrations, the values of the supply voltages, and other system parameters. Since in (39) the squares of the damping factors are subtracted, it is evident that their influence is also small because at close values their difference is zero.

In order to perform an approximate sensitivity analysis of the method, Formula (39) is transformed into the form

$$f_{cusp} = \frac{\sqrt{2}}{2} \sqrt{\frac{f_{s1}^2 + f_{s2}^2}{1 - \frac{2(\eta_1^2 - \eta_2^2)}{f_{s1}^2 - f_{s2}^2}}}.$$
(40)

It is assumed that due to the same geometrical and physical parameters, both beams have almost the same losses, i.e.,

η

$$_1 \approx \eta_2$$
 (41)

and Expression (40) can be assumed to be approximately equal to

$$f_{cusp} \approx \frac{\sqrt{2}}{2} f_{s1} \sqrt{1 + f_{s21}^2},$$
 (42)

where

$$f_{s21} = \frac{f_{s2}}{f_{s1}}.$$
(43)

When the natural frequency f_{s1} of microcantilever 1 increases by a small value Δf_{s1} , the new cusp point is calculated by the formula

$$f_{cusp1} \approx \frac{\sqrt{2}}{2} (f_{s1} + \Delta f_{s1}) \sqrt{1 + f_{s21}^2};$$
 (44)

where the ratio f_{s21} is assumed to be independent on Δf_{s1} due to its small value. Considering (40), the change in the cusp point frequency value in this case is

$$f_{cusp} - f_{cusp1} \approx \frac{\sqrt{2}}{2} \Delta f_{s1} \sqrt{1 + f_{s21}^2},$$
 (45)

from which the relationship between the frequency change Δf_{s1} and the cusp point change is obtained:

$$\Delta f_{s1} \approx \frac{\sqrt{2}(f_{cusp} - f_{cusp1})}{\sqrt{1 + f_{s21}^2}},$$
(46)

considering that the natural frequency of microcantilever 1 can be represented by the expression

$$f_{s1} = \frac{1}{2\pi} \sqrt{\frac{k_{1e}}{m_{1e}}},$$
(47)

where k_{1e} and m_{1e} are the effective stiffness and effective mass of cantilever 1, respectively. It follows from (47) that

$$m_{e1} = \frac{k_{1e}}{2\pi^2 f_{s1}^2},\tag{48}$$

Assuming that the effective mass has increased by a small value Δm_1 , the new mass is as follows:

$$m_{e1} + \Delta m = m_{e11} = \frac{k_{1e}}{4\pi^2 f_{s11}^2},\tag{49}$$

where f_{s11} is the natural frequency effective variation in microcantilever 1 due to the added mass. By (48) and (49), the difference of the squares of the natural frequencies can be expressed as

$$f_{s1}^2 - f_{s11}^2 = \frac{k_{1e}}{4\pi^2} \left(\frac{1}{m_{e1}} + \frac{1}{m_{e11}} \right).$$
(50)

The resulting expression is transformed after the simplifications

$$f_{s1}^2 - f_{s11}^2 = (f_{s1} + f_{s11})(f_{s1} - f_{s11}) \approx 2f_{s1}(f_{s1} - f_{s11}),$$
(51)

are made, because it is assumed that $f_{s1} \approx f_{s11}$ and

$$\frac{1}{m_{e1}} - \frac{1}{m_{e11}} = \frac{m_{e11} - m_{e1}}{m_{e1}m_{e11}} \approx \frac{\Delta m_{e1}}{m_{e1}^2}.$$
(52)

Here, the simplification is based on the small differences of masses, i.e., $m_{e1} \approx m_{e11}$. Once the above simplifications are made, the difference in natural frequencies is found:

$$\Delta f_{s1} = f_{s1} - f_{s11} \approx \frac{k_{1e} \Delta m_{e1}}{8\pi^2 m_{e1}^2 f_{s1}} = \frac{f_{s1} \Delta m_{e1}}{2m_{e1}}.$$
(53)

From (46) and (53), the relationship between the cusp points frequency variation and the effective mass of the microcantilever 1 is given by

$$\Delta m_{e1} \approx \frac{2\sqrt{2}(f_{cusp} - f_{cusp1})m_{e1}}{\sqrt{f_{s1}^2 + f_{s2}^2}}.$$
(54)

Using this expression, one can calculate what the resolution of the method is; for example, with a measurement limit of $\Delta f_{cusp_{\min}} = (f_{cusp} - f_{cusp1})_{\min} = 0.01$ Hz, an effective mass of $m_{e1} = \frac{33m_1}{280} = 4.43951 \times 10^{-11}$ kg [23], and beam natural frequencies according to Table 1, a limiting sensitivity lower than 1.5×10^{-17} kg or 15 fkg is obtained. Here, the mass of the microcantilever is 14.108×10^{-10} kg. From Equation (54), it is concluded that to increase the limiting sensitivity, it is necessary to improve the measurement accuracy and increase the natural frequencies of the microcantilevers. At natural frequencies of the microcantilevers twice as high, for the considered case, the limit resolution is increased by one order of magnitude.

7. Experimental Determination of the Capabilities of the Method, Changing One of the Natural Frequencies of the Microcantilevers by Heating

The experimental setup shown schematically in Figure 8 was used to investigate the sensitivity of the detection method. The heater of microcantilever 1 was connected to a battery via a serially connected ammeter and a variable resistor. The current in the heater was continuously adjusted through the adjustable resistor, which caused the microcantilever to heat up at different temperatures. As a result of the increased temperature, the microcantilever changed its dimensions in proportion to the coefficient of thermal expansion (CTE), and the natural frequency changed in proportion to a parameter called the temperature coefficient of frequency (TCF) [31–33].



Figure 8. A schematic of the experiment to change the temperature of microcantilever 1 by Joule heating.

The experiments were conducted after the common base of the two microcantilevers was vibrationally driven by setting 400 uniformly varying values in the range of the two natural frequencies of the microcantilevers. The vibration time of each of these frequencies was selected to be greater than five oscillation periods. For each step of these 400 frequencies, the electrical voltages of the two half-bridges were measured and processed according to the methodology described above. The electric current in the heater of microcantilever 1 was varied from 0 to 1800 μ A by setting 20 different values. Twenty Technical Data Management (TDM) files [34] were generated using LabVIEW 11 with the recorded values of the differences of the voltage amplitude–frequency characteristics of the two half-bridges. The TDM files were converted to an Excel file and then processed using the Maple program. The Maple program algorithm finds the smallest value of the absolute voltage V_{abs} array

and divides it into left $V_{absleft}$ and right $V_{absright}$, and then approximates them with the parabolas, respectively,

$$\widetilde{V}_{absleft} = a_l + b_l f + c_l f^2
\widetilde{V}_{absright} = a_r + b_r f + c_r f^2$$
(55)

The intersection f_{ce} of these parabolas is called the experimental cusp point. The processing of the experimental results for a single file is illustrated graphically in Figure 9.



Figure 9. Experimental data processing for V_{abs} from an Excel file obtained at current $i = 1053 \mu$ A. The frequency of the forced vibrations at the base of the two microcantilevers was varied in the range [65,520, 66,150] Hz.

Figure 10a illustrates the obtained experimental relationship between the cusp point frequencies f_{ce} and the values of the heating current *i*. The experimental data are approximated linearly by a line of the type $f_{capr} \approx bi + c$, depicted in Figure 10a with a solid line. The natural frequency of microcantilever 1 is expressed by (39) and the results with their corresponding linear approximation are plotted in Figure 10b.



Figure 10. Experimental data on the effect of heating microcantilever 1 on the frequency of the experimental cusp point, natural frequency, and their approximation: (**a**) the cusp point frequency as a function of the heater current and its approximating linear relationship; (**b**) tge dependence of the natural frequency of microcantilever 1 on its heating current *i* and the approximating line.

8. Determine the Sensitivity of the Detection Method by Examining the Offset of the Cusp Point

In order to investigate the sensitivity of the sensor with respect to the mass added to microcantilever 1, we assume here the addition of a thought uniform homogeneous layer of mass Δm_{scv} of the same density as that of the beam material to the active surface of the microcantilever. Then, the thickness l_3 of the beam will increase by Δl_3 as the added mass Δm_{scv} is calculated by the formula

$$\Delta m_{SCV} = \rho l_1 l_2 \Delta l_3,\tag{56}$$

where

$$\Delta l_3 = \frac{\Delta m_{SCV}}{\rho l_1 l_2}.$$
(57)

The natural frequency of microcantilever 1 with the additional layer according to [35–38] is calculated by

$$f_{1SCV} = \frac{1.875104069^2}{2\pi} \sqrt{\frac{EI_{SCV}}{\rho A_{SCV} l_1^4}},$$
(58)

whereby replacing

$$I_{SCV} = \frac{l_2(l_3 + \Delta l_3)^3}{12},$$
(59)

$$A_{SCV} = l_2 (l_3 + \Delta l_3), \tag{60}$$

and taking (57) into account, after transformations, the following is obtained:

$$f_{1SCV} = 0.161540067953243 \frac{m + \Delta m_{scv}}{l_1^3 l_2 \rho} \sqrt{\frac{E}{\rho}},$$
(61)

where $m = l_1 l_2 l_3 \rho$ is the mass of microcantilever 1 without the added thought layer. Here, the calculations are performed with double precision to reflect small masses on the order of femtograms.

If the above formula is substituted with $\Delta m = 0$, one arrives at the simplification

$$f_{s1} = 0.1615400679053 \frac{l_3}{l_1^2} \sqrt{\frac{E}{\rho}},\tag{62}$$

which is the natural frequency of the unchanged microcantilever 1.

In order to investigate the correlation between the added mass and the frequency, the frequency difference was found:

$$\Delta f_1 = f_{1SCV} - f_{s1} = 1.609414378345883929 \times 10^{14} \Delta m_{scv}, \tag{63}$$

from which the inverse dependence follows

$$\Delta m_{scv} = 6.213440201943362337853559 \times 10^{-15} \Delta f_1.$$
(64)

Figure 11 shows the equivalent change in mass that would have resulted from the temperature change of the natural frequency of microcantilever 1, illustrated by Figure 10b. In the figure, the processed experimental data are represented by an asterisk symbol, and the approximated dependence according to Equation (64) is plotted as a solid straight line.

By the coefficients in front, Δm_{scv} and Δf_1 are the frequency sensitivities from the mass and vice versa, respectively. By (63) and (64), it is calculated that when the experimental system can be measured to the nearest 0.1 [Hz], it will register a mass that, for every hundredth of a Hz, increases by 6.21344 × 10⁻¹⁶ kg. This corresponds to 6.21344 × 10⁵ fg. The mass of a SARS-CoV-2 virus is on the order of 1 fg = 1 × 10⁻¹⁸ kg. One person was found to carry 10^{10} to 10^{11} viruses with a total mass of 1–100 µg during the peak of infection [39,40].





The result obtained here indicates that with the accuracy of the measurement system and the adopted parameters of the microcantilevers thus established, it will be possible to detect SARS-CoV-2 viruses at the initial stage of infection before infection occurs. The sensitivity of the system allows it to be applied to the detection of other viruses, pathogens, and chemical substances. There is some margin for improvement in sensitivity in the accuracy of the system itself. Increasing the natural frequency of the microchannels will also have a beneficial effect on sensitivity. This study showed that of the two microcantilevers, the one with the higher natural frequency has a higher sensitivity, and hence, it is advisable to keep this one active.

9. Conclusions

The theoretical foundations of a new method for the detection of objects with masses on the order of femtograms are described and justified. Although sensors with two microcantilevers have a well-known structure with one active and another passive microcantilever, and their investigation methods are well known, a new method is proposed here that allows us to sum the amplitude–frequency responses of the voltages at the two Wheatstone half-bridges, thus avoiding the influence of the phase shift of the signals from the piezoresistive sensors in the microcantilevers. Another novelty is the measurement of the cusp point frequency of the amplitude–frequency response of the differences of the two half-bridge voltages, which provides higher accuracy compared to offset-based eigenfrequency methods.

A formula was derived that gives the relationship between the cusp point frequency and the natural frequencies of the microcantilevers. The relationship between the cusp points frequency variation and the effective mass of microcantilever 1 was obtained. The analytical relationships for the variation in the mass of microcantilever 1 and the cusp point frequency were derived, which also determine the sensitivity of the method.

A high-precision experimental system was designed to investigate the method, by which controlled harmonic excitation was generated at the common base of the microcantilevers, and simultaneously, the voltages obtained from piezoresistive sensors formed on the microcantilevers were measured. Using a LabVIEW 11 program in real time, the excitation of the base was simultaneously controlled, and the result obtained due to the deformations of the piezoresistive sensors was measured. The experimental system combined with the proposed measurement method allowed the measurement of the cusp point frequency in the amplitude–frequency response with a resolution of 0.01 Hz, which is sufficient to detect the presence of SARS-CoV-2 virus at an early stage before infection has occurred.

The method was verified by a temperature shift of the frequency of microcantilever 1, and the offsets of the natural frequency and the magnitude of the equivalent femtogram mass that would cause it were determined.

The dual-microcantilever piezoresistive sensor with the experimental system described here and the measurement method applied are universal in nature and can find various applications in medical, chemical, environmental, and other research.

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